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To: John Lohmeyer, Chairman, X3T10
X3T10/94-217r0
Copy: SCSI-3 WG and interested parties
From: Kevin Gingerich, Texas Instruments, Inc.
Subj: Proposed Implementor's Note for Device Capacitance and Spacing for Fast-20 SCSI
The SCSI bus is a distributed parameter circuit whose electrical characteristics and responses are primarily defined by the distributed inductance and capacitance along the physical media. The media is defined here as the interconnecting cable(s) or conducting paths, connectors, terminators, and SCSI devices added along the bus. The following analysis derives a guideline for the amount of capacitance (and its spacing) that can be added to the Single-ended Fast-20 SCSI bus.

To a good approximation, the characteristic transmission line impedance seen into any cut point in the unloaded SCSI bus is defined by $Z=\sqrt{L / C}$, where $L$ is the inductance per unit length and $C$ is the capacitance per unit length. As capacitance is added to the bus, in the form of devices and their interconnection, the bus impedance is lowered and can be expressed by $Z^{\prime}=\sqrt{L /\left(C+C^{\prime}\right)}$, where $C^{\prime}$ is the added capacitance per unit length. When capacitance is added to the bus by devices, an impedance mismatch occurs and when a signal wave arrives at this mismatch in impedance, an attenuation (or amplification) of the signal will occur. The magnitude of the attenuation will depend upon the ratio of the mismatched impedance or $A=Z^{\prime} / Z$, where $Z$ is the load impedance and $Z$ is the source impedance.

Substituting the equations for $Z$ and $Z$ and reducing,

1) $\quad A=Z^{\prime} / Z=\frac{\sqrt{L /\left(C+C^{\prime}\right)}}{\sqrt{L / C}}=\sqrt{\frac{1}{\left(1+C^{\prime} / C\right)}}$.

We now have a relationship for the attenuation of the signal voltage at an impedance mismatch due to load capacitance distributed on the SCSI bus. Next, a rule for the ratio of $Z$ to $Z$ will be derived.

With fast transfer rates and electrically long ${ }^{1}$ media, it becomes essential to achieve a valid input voltage level on the first signal transition from an output driver anywhere on the bus. This is called incident-wave switching. If inci-dent-wave conditions are not achieved, reflected-wave switching must be used. Reflected-wave switching depends upon reflected energy occurring some time after the first transition arrives to achieve a valid logic voltage level. In the Fast-20 SCSI environment, the valid low-level input voltage threshold has been raised and the high-level input voltage threshold has been lowered to allow incident-wave switching with some inevitable impedance mismatching and signal attenuation along the media.

The signal voltage at an impedance mismatch is $V_{L 1}=V_{L 0}+V_{J 1}+V_{R 1}$, where $V_{L 0}$ is the initial voltage, $V_{J 1}$ is the input signal voltage, and $V_{R 1}$ is the reflected voltage. The voltage reflected back from the mismatch is $V_{R 1}=\rho\left\llcorner\times V_{J 1}\right.$
${ }^{1}$ Electrically long is defined here as $\tau>\frac{t_{10-90 \%}}{3}$, where $\tau$ is the one-way time delay across the bus and $t_{10-90 \%}$ is the $10 \%$ to $90 \%$ transition time of the fastest driver output signal.
where, $\rho\left\llcorner=\frac{Z^{\prime}-Z}{Z^{\prime}+Z}\right.$ and is the coefficient of reflection commonly use in transmission line analysis. The voltage equation can now be written as $V L 1=V \angle 0+V J 1+\rho L \times V_{J 1}$.

When a SCSI signal is asserted, the $V_{L 0}$ can be at a maximum of 3.7 V and go to 0 V (for a perfect driver) giving a $V_{J 1}$ of -3.7 V and the signal voltage must go below the minimum receiver input voltage threshold of 1 V . In equa-$1>3.7+(-3.7)+\rho\llcorner\times(-3.7)$
tion form, $\rho_{\llcorner }>\frac{1-3.7+3.7}{-3.7}=-0.27$
age can be reflected back towards the source or the minimum assertion level will not be achieved by the incident wave ${ }^{2}$.

Now, to relate this to $Z^{\prime} / Z$ and using equation 1) for $C / C$,

$$
\rho_{\mathrm{L}}=\frac{Z^{\prime}-Z}{Z^{\prime}+Z}>-0.27
$$

2) 

$$
Z^{\prime}-Z>-0.27\left(Z^{\prime}+Z\right)
$$

$$
Z^{\prime}(1+0.27)>Z(1-0.27)
$$

$$
\frac{Z^{\prime}}{Z}>\frac{0.73}{1.27}=0.57
$$

$$
\begin{aligned}
& 0.57<\sqrt{\frac{1}{\left(1+\frac{C^{\prime}}{C}\right)}} \\
& \text { and } \frac{1}{0.57^{2}}-1>\frac{C^{\prime}}{C} \\
& \frac{C^{\prime}}{C}<2.08
\end{aligned}
$$

. The negative value means that no more than $27 \%$ of the input signal volt-

We can now say that capacitance should not be added at more than twice the bus distributed capacitance for incident wave switching. For example, a cabled bus with $L=295 \mathrm{nH} / \mathrm{m}(90 \mathrm{nH} / \mathrm{ft})$ and $C=41 \mathrm{pF} / \mathrm{m}(12.5 \mathrm{pF} / \mathrm{ft})$ and $Z=85 \Omega$, the guideline becomes to add no more than $85 \mathrm{pF} / \mathrm{m}(26 \mathrm{pF} / \mathrm{ft})$ anywhere along the bus. This guideline can be met by 25 pF loads spaced $0.3 \mathrm{~m}(1 \mathrm{ft})$ from each other, 50 pF spaced $0.6 \mathrm{~m}(2 \mathrm{ft})$ apart, or 12.5 pF spaced $0.15 \mathrm{~m}(0.5 \mathrm{ft})$ apart. This relationship is shown graphically in figure 1.

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Figure 1. Minimum device spacing versus bus and device capacitance.


[^0]:    ${ }^{2}$ A similar analysis can be used for the negation case of 0 V to $2.8 \mathrm{~V}([48 \mathrm{~mA}+22 \mathrm{~mA}] \times 40 \Omega)$ and an input voltage threshold of 1.9 V for a minimum reflection coefficient of -0.32 . This leaves assertion as the most restrictive case.

