

00-104R0:
Ultra-320 SCSI Compensation Techniques

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- We will collect more data on different configurations:
 - heavily and lightly loaded busses
 - ✓• typical and atypical
 - point-to-point
- We will investigate:
 - ✓• can lower amplitudes be used to address large chip power requirements?
 - ✓• how bad will common-mode degrade for large amplitudes?
 - ✓• would receiver compensation work?
 - ✓• how much capacitance will be added by larger drivers?
 - how much capacitance is acceptable?
 - could we use a different terminator scheme?

Part I: Pre-compensation with XTALK

- Further investigate write pre-compensation for Ultra-320.
- Estimate amplitude and timing factors to define eye mask:
 - Clocking;
 - De-skew.
- Measure the signal degradations to find eye opening with ISI and reflections for typical configurations:
 - Amplitude noise;
 - Timing shift;
 - Miscellaneous noise.
- Measure the signal degradations with XTALK as well.
- How much does pre-comp aid reception:
 - Setup margin;
 - Amplitude margin.

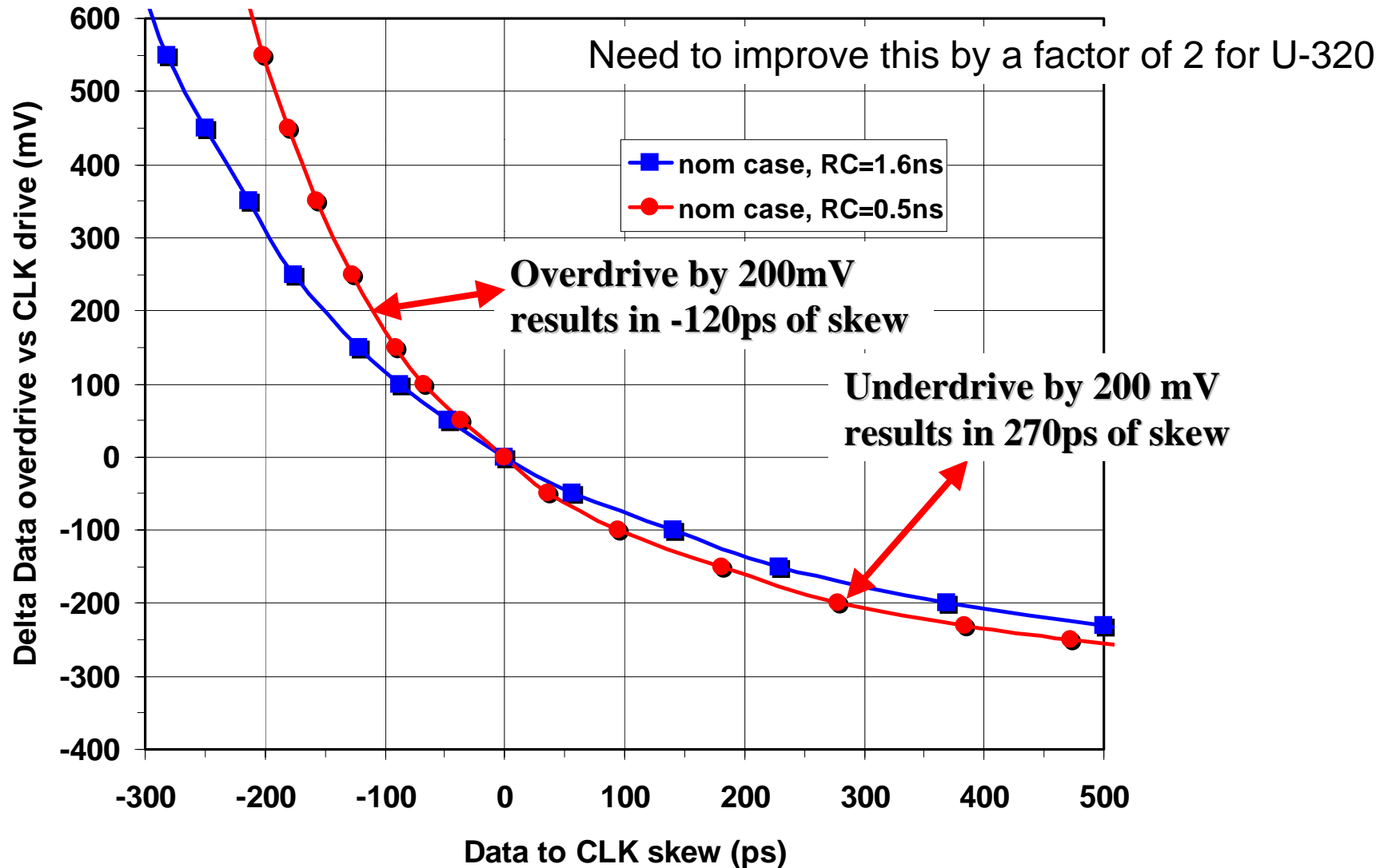
Random amplitude (0-to-pk)	Terminator voltage mismatch	(SPI-3: Tab20)	13mV
	Terminator resistance mismatch	(SPI-3: Tab20)	5mV
	Driver error	(SPI-3: TabA2)	40mV
	Receiver comparator	(SPI-3: TabA5)	30mV
	Root sum squares of random amplitude		52mV
Deterministic Amplitude: (0-to-pk)	Cable + back-plane resistance (cable spec + meas)		28mV
	Comparator overdrive requirement	(SPI-3: Fig48)	70mV
	Total amplitude 0-to-pk factors:		150mV
Timing factors (0-to-pk)	Low Vt vs substrate noise	(99-261r1)	100ps
	Receiver clock jitter	(99-261r1)	125ps
	Residual de-skew	(99-261r1)	125ps
	De-skew stability	(temperature)	100ps
	Input slew rate dependent skew	(99-261r1)	100ps
	Receiver amp dependent delay	(99-261r1 [*])	150ps
	Receiver FF rise/fall prop delay difference		300ps
	Total 0-to-pk timing factors:		1.0ns

These are the error sources that are not accounted for by our test setup as well as those in the SPI-3 budget. 0-to-pk values converted from pk-to-pk numbers by a factor of 1/2.

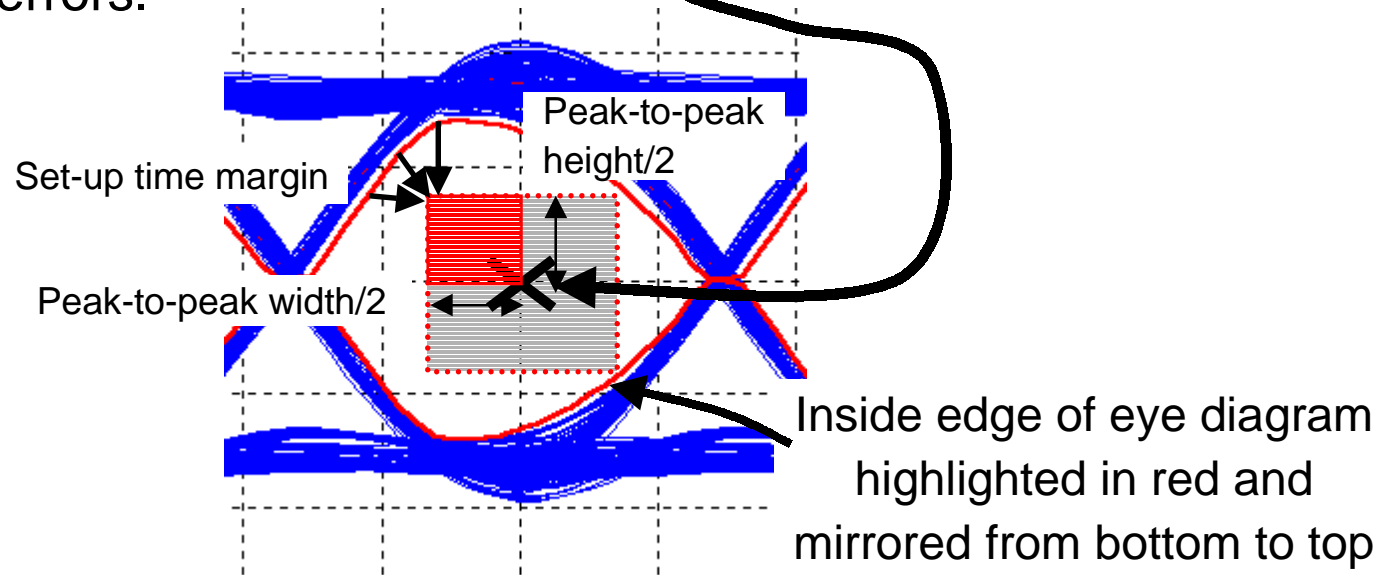
* refer to diagram on next page.

Typical U-160 Receiver Amplitude Dependent Delay

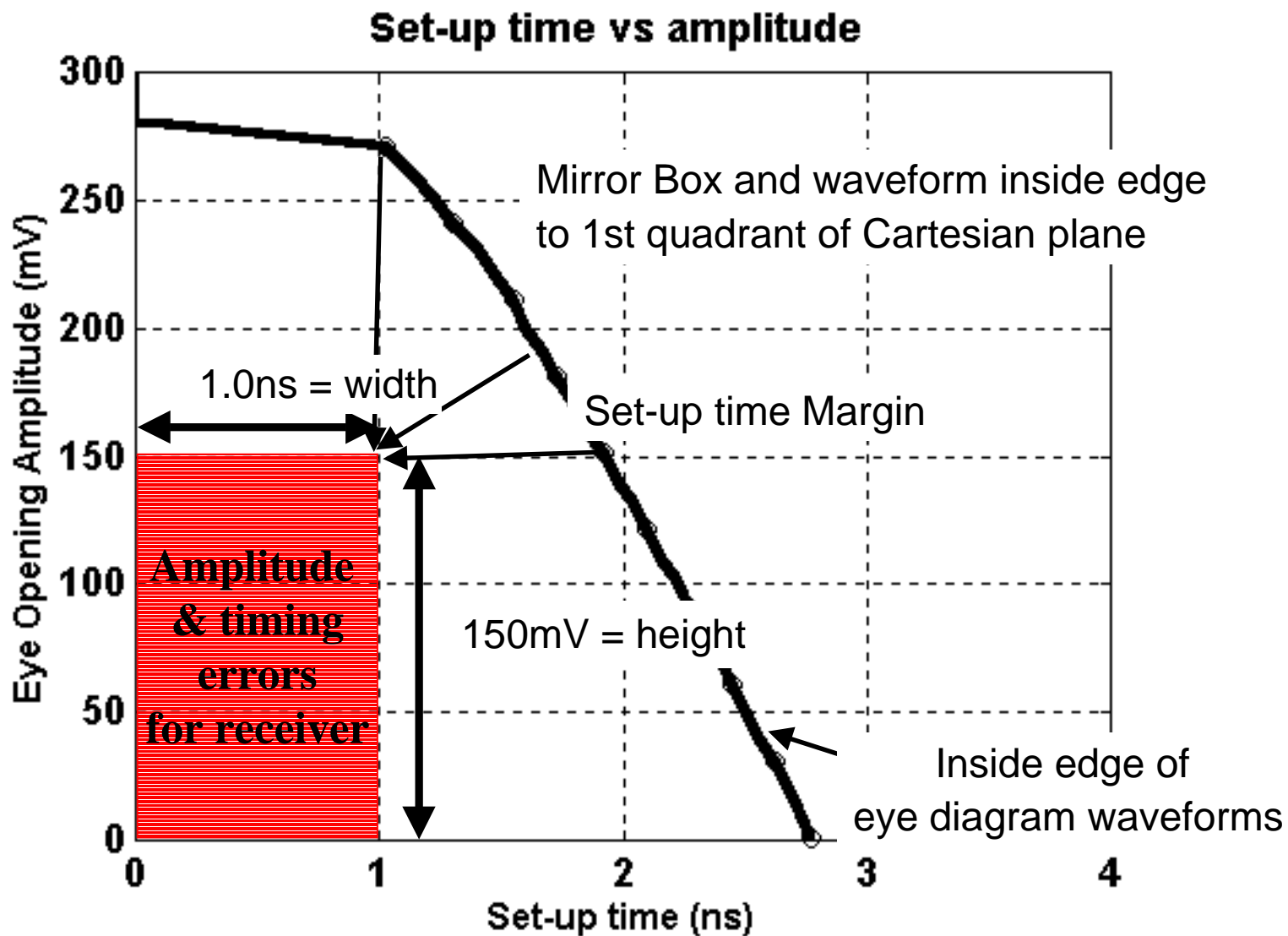
Data to CLK Skew versus Overdrive



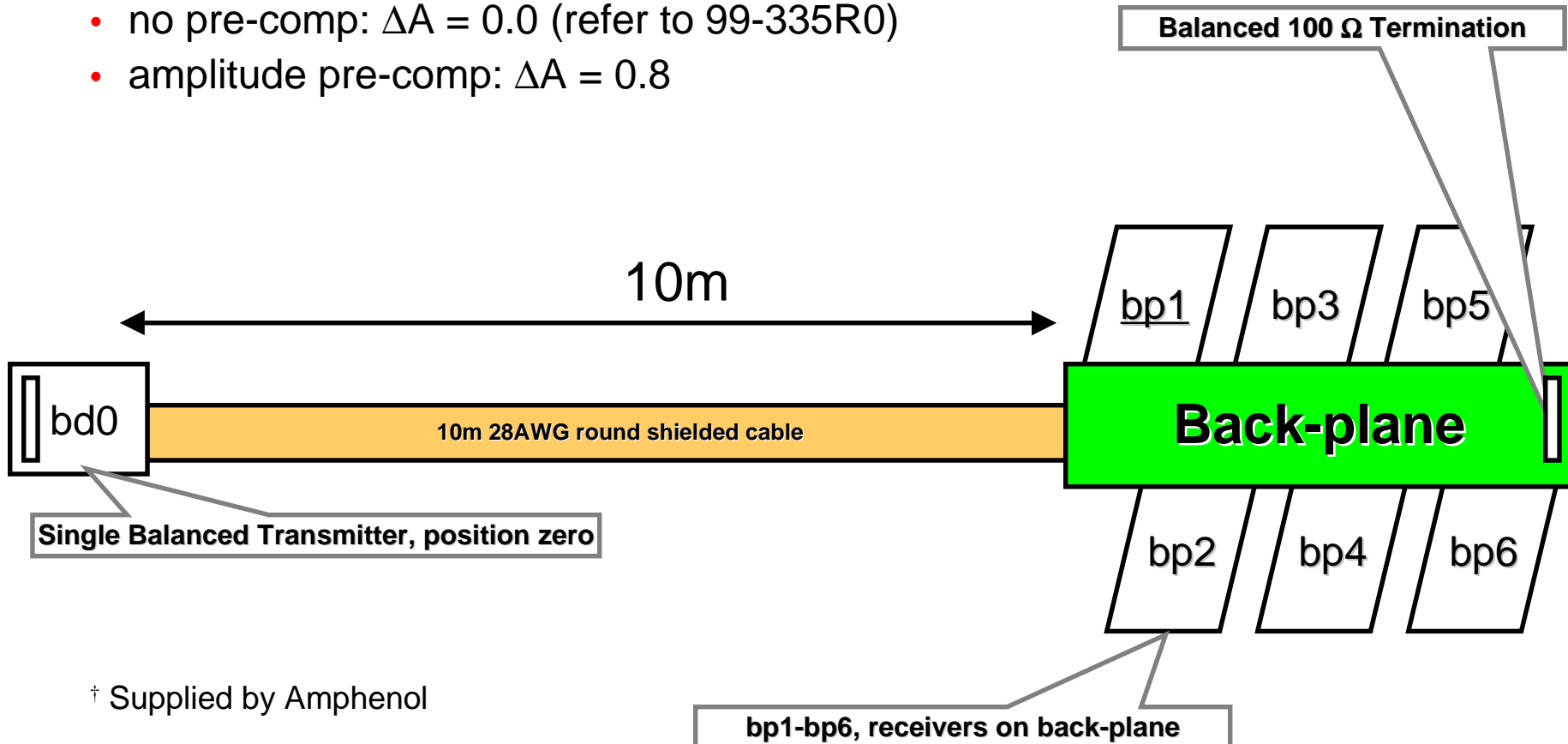
- Error sources are used to define range over which receiver characteristic may typically vary from the ideal sample point. I.e. actual sample point may lie anywhere within dashed box defined by $2 \times$ 0-to-pk height and $2 \times$ 0-to-pk width of errors.



- Amplitude error sources define height & Timing error sources define width.
- E.g. set-up time margin is measured as distance from eye diagram waveform to box.
- Mirror top/bottom waveforms and box to first quadrant to visualize margin.
- Only concerned with set-up time in this presentation.

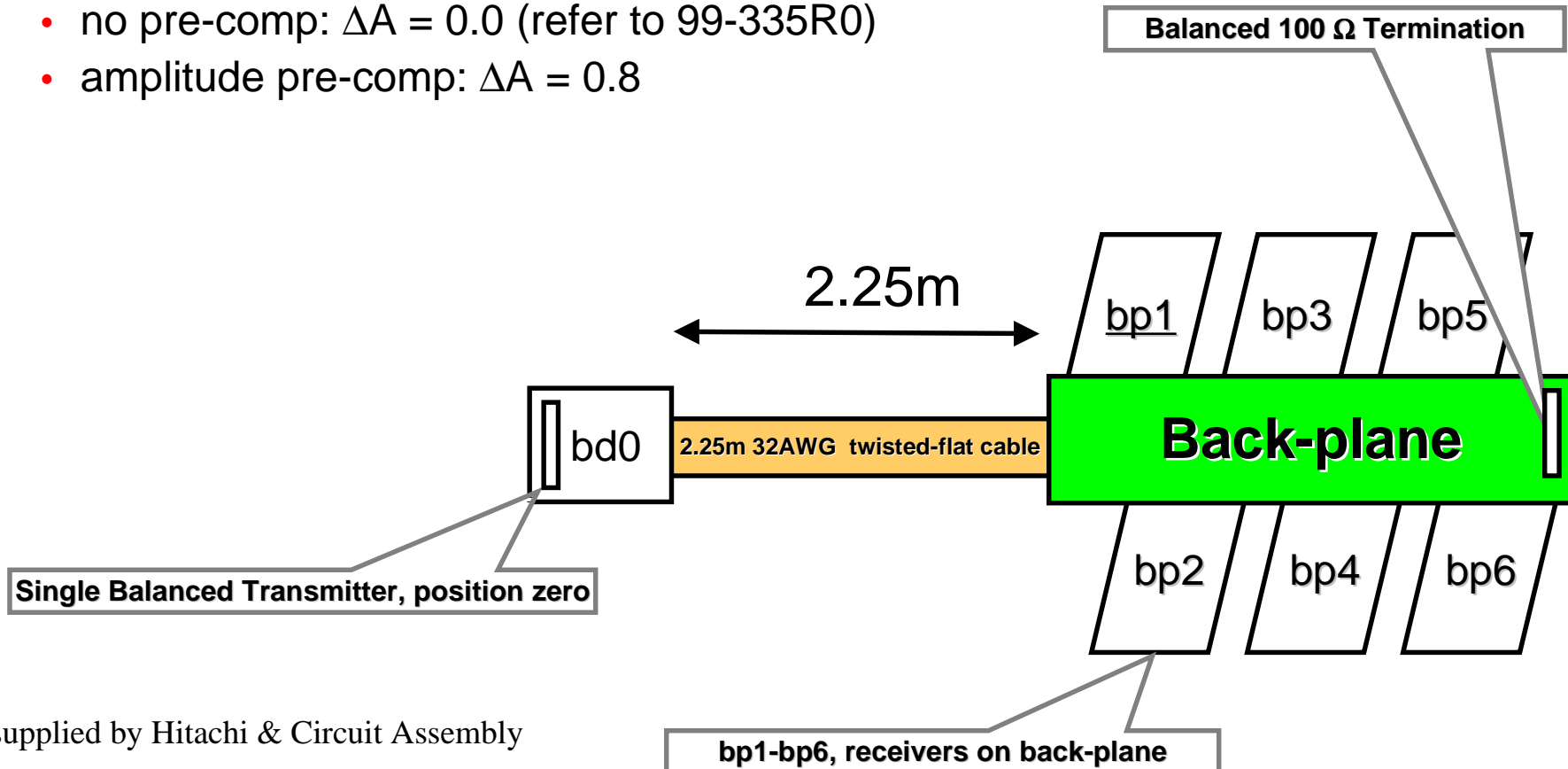


- 10 meter, Madison 28AWG[†] round shielded cable, plus 6-slot back-plane.
- Waveforms captured @ 4Gs/s:
 - no pre-comp: $\Delta A = 0.0$ (refer to 99-335R0)
 - amplitude pre-comp: $\Delta A = 0.8$

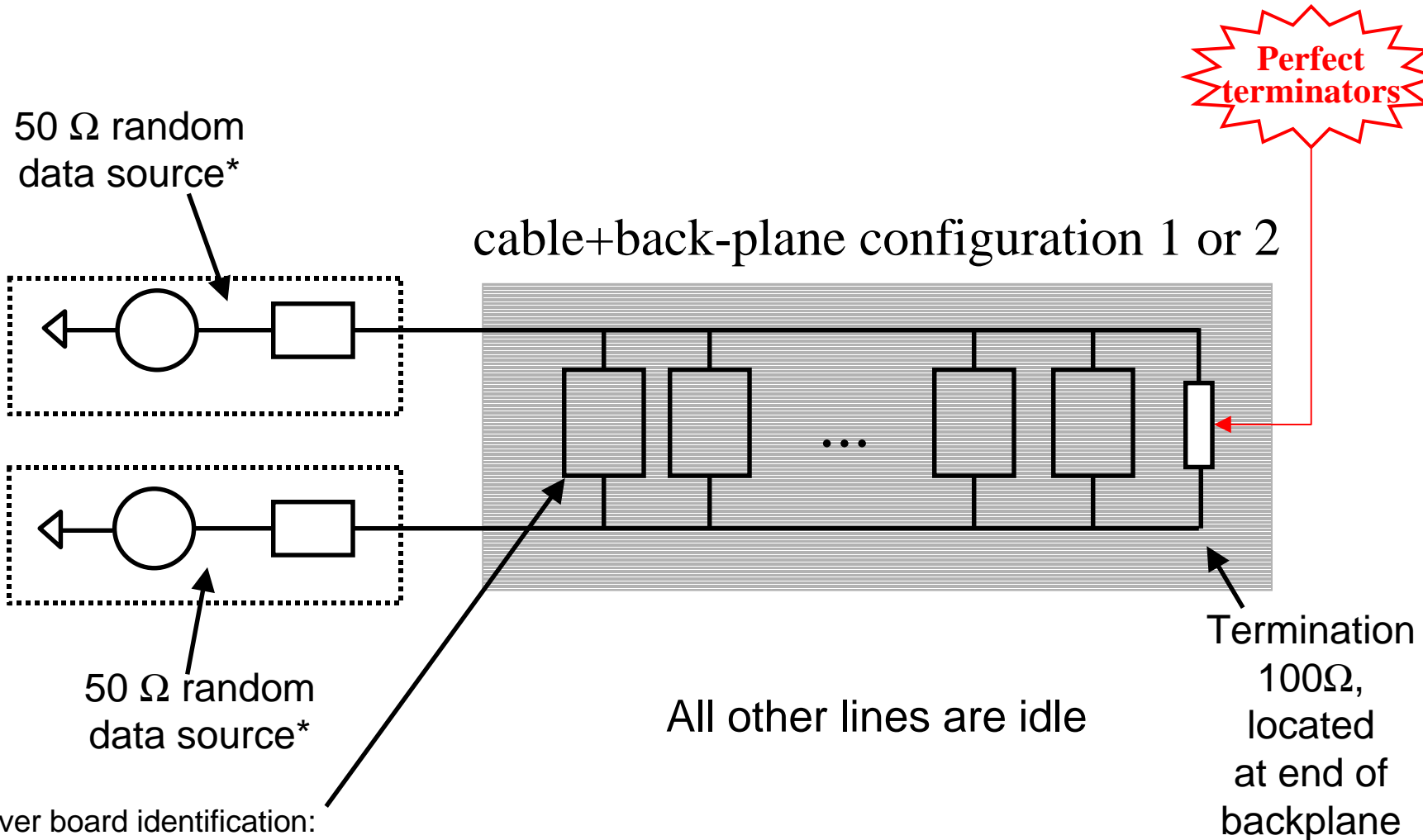


[†] Supplied by Amphenol

- 2.25 meter, Hitachi 32AWG twisted-flat cable[†], plus 6-slot back-plane.
- Waveforms captured @ 4Gs/s:
 - no pre-comp: $\Delta A = 0.0$ (refer to 99-335R0)
 - amplitude pre-comp: $\Delta A = 0.8$

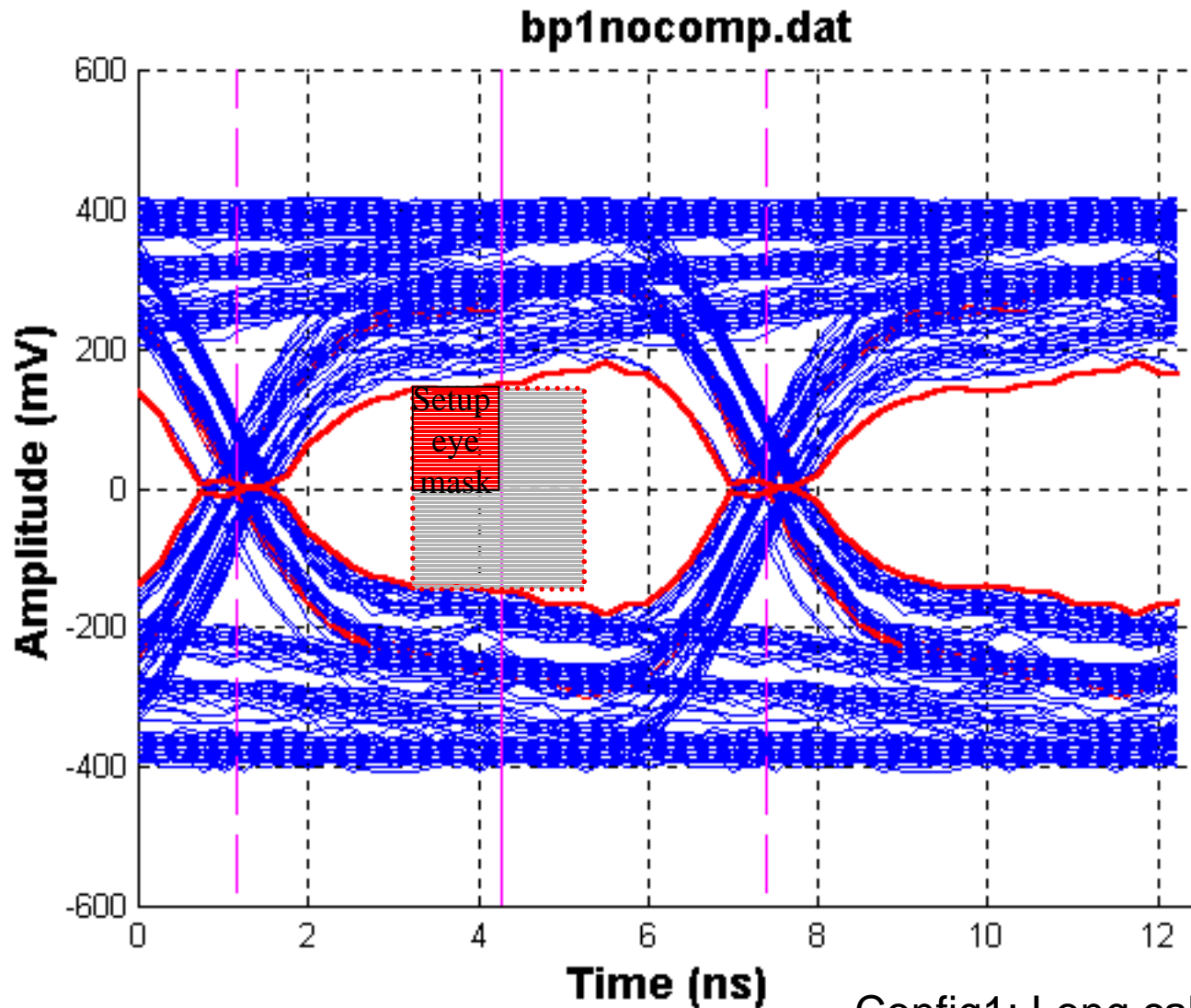


[†]supplied by Hitachi & Circuit Assembly

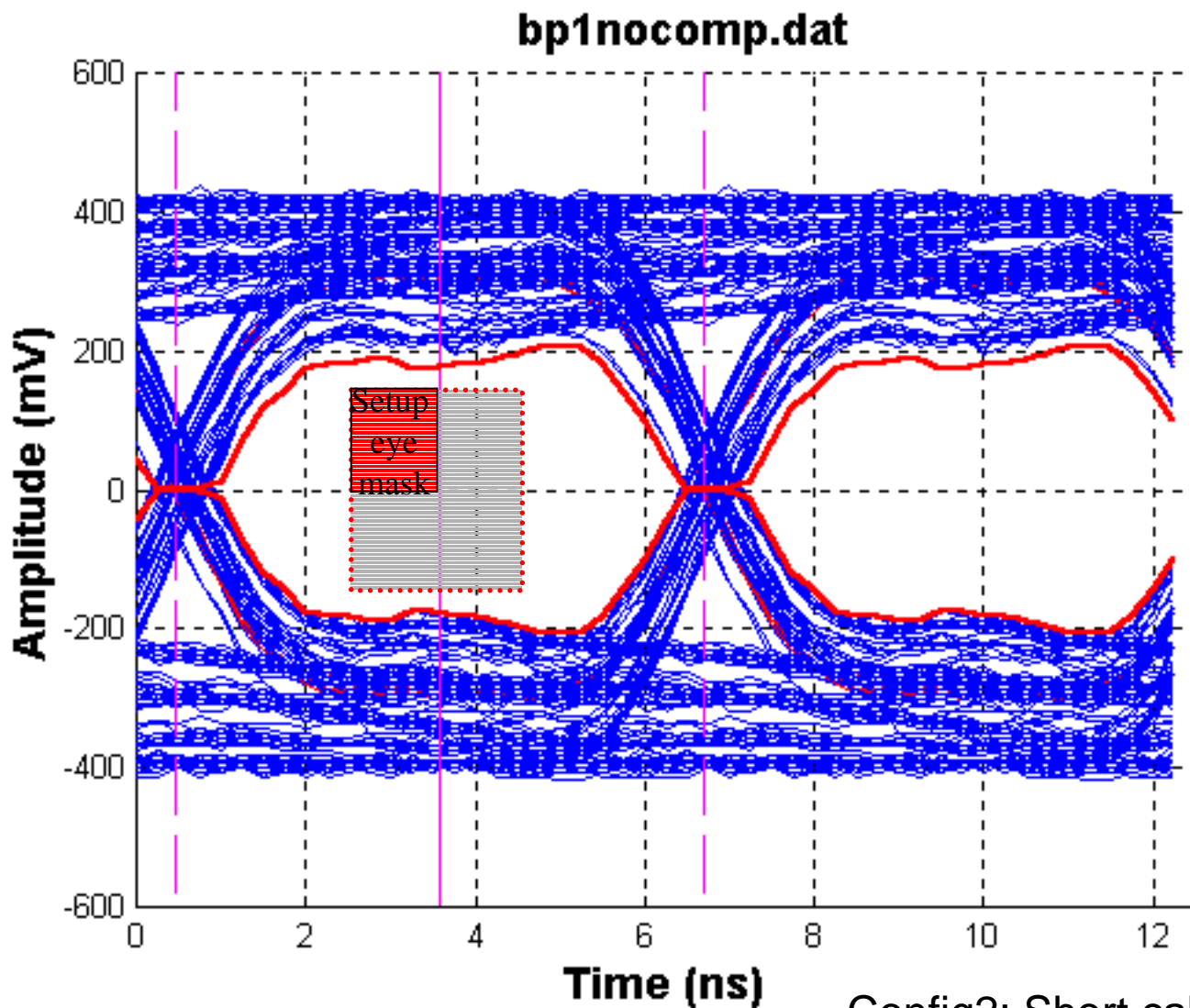


receiver board identification:
-bpN are on the back-plane
-bdN are on connectors along cable

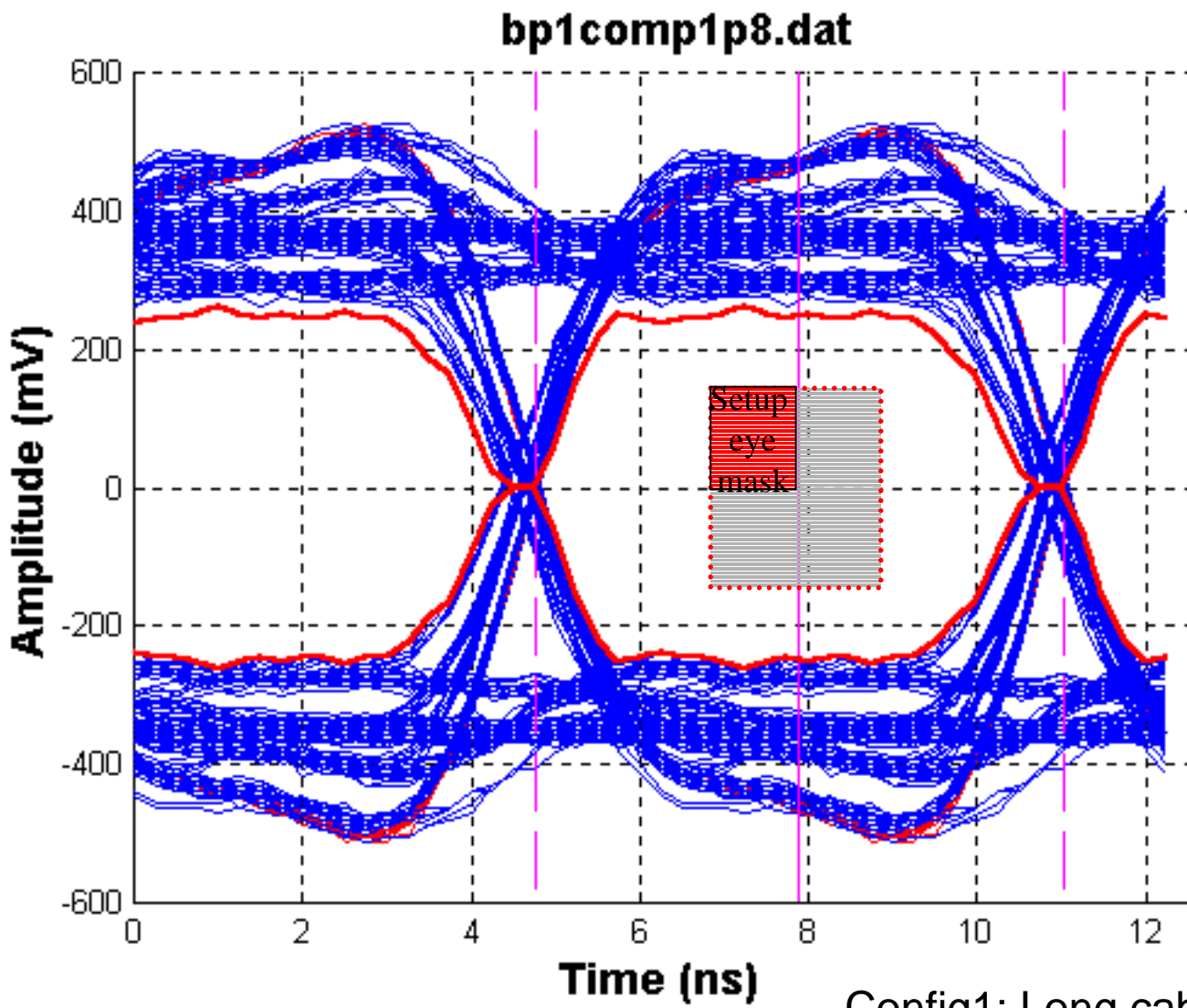
* TEK 2041



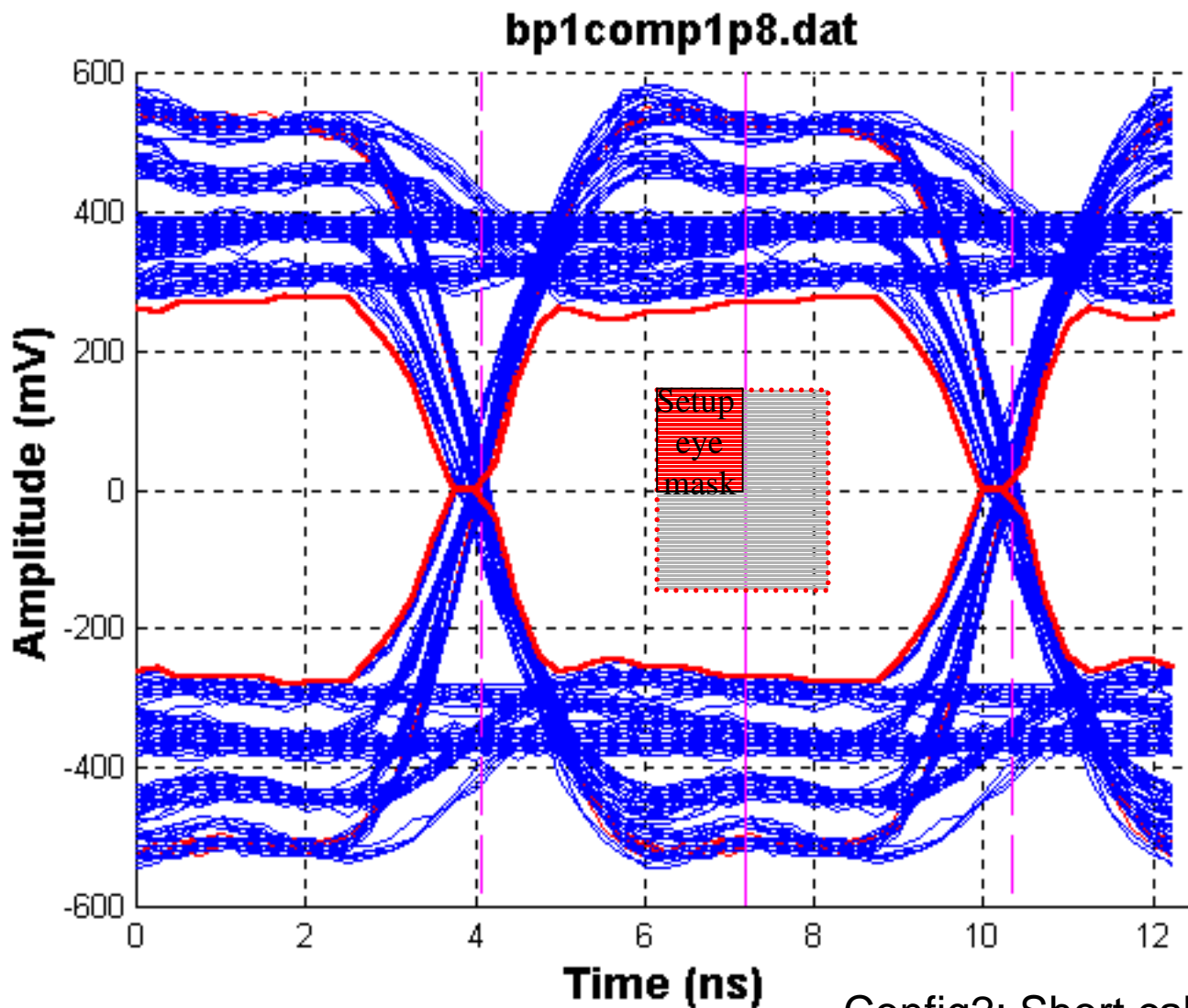
Config1: Long cable + back-plane



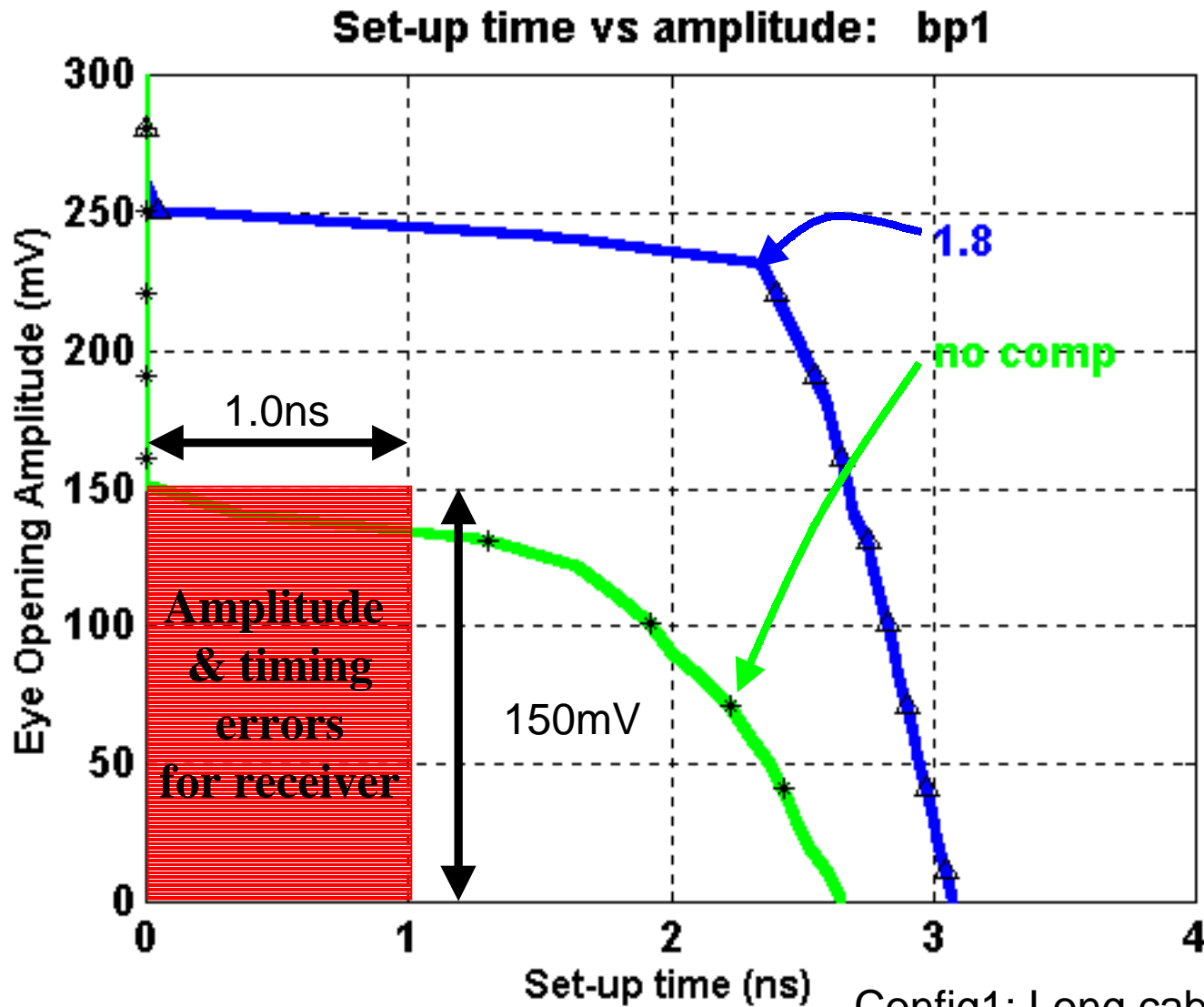
Config2: Short cable + back-plane



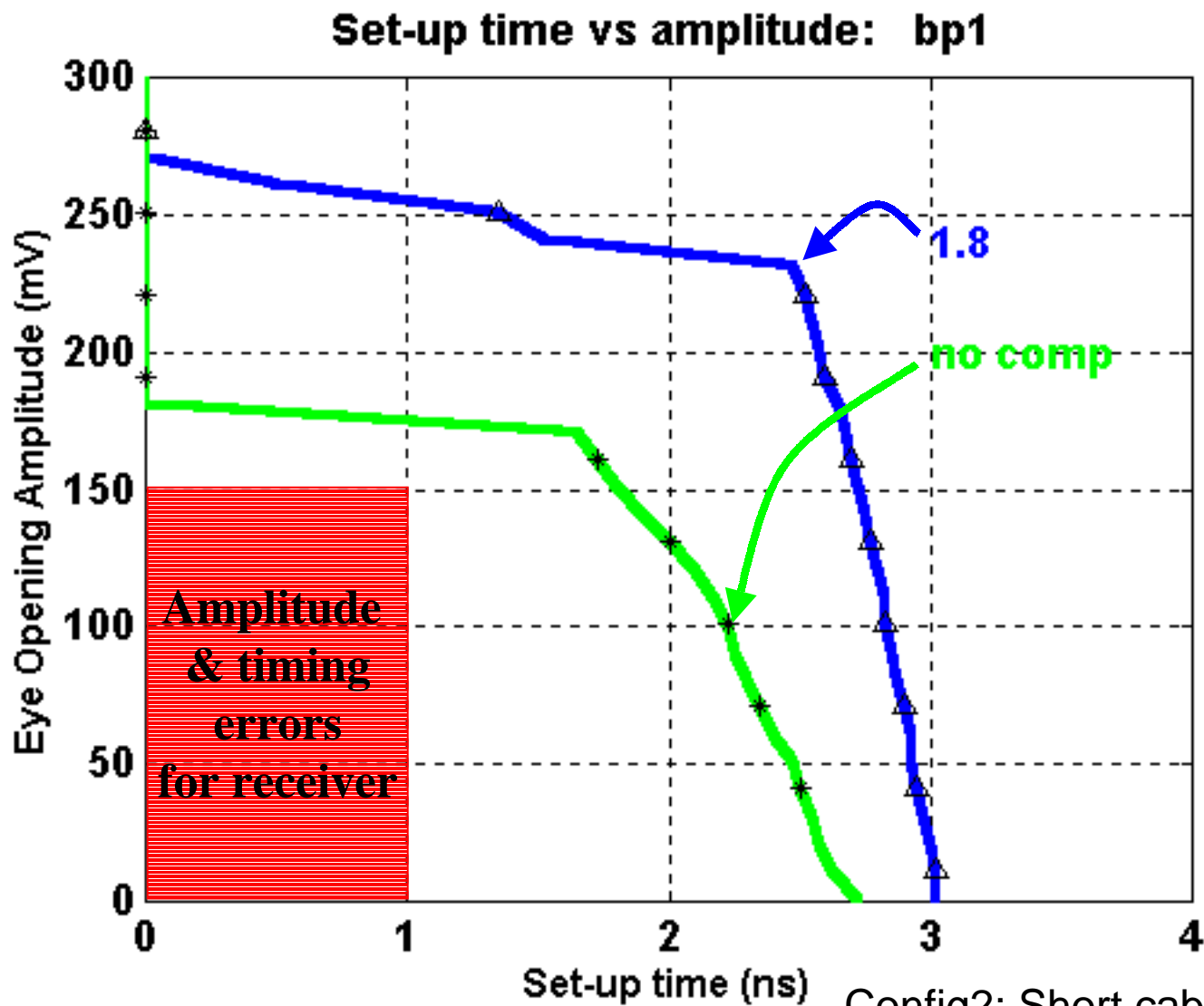
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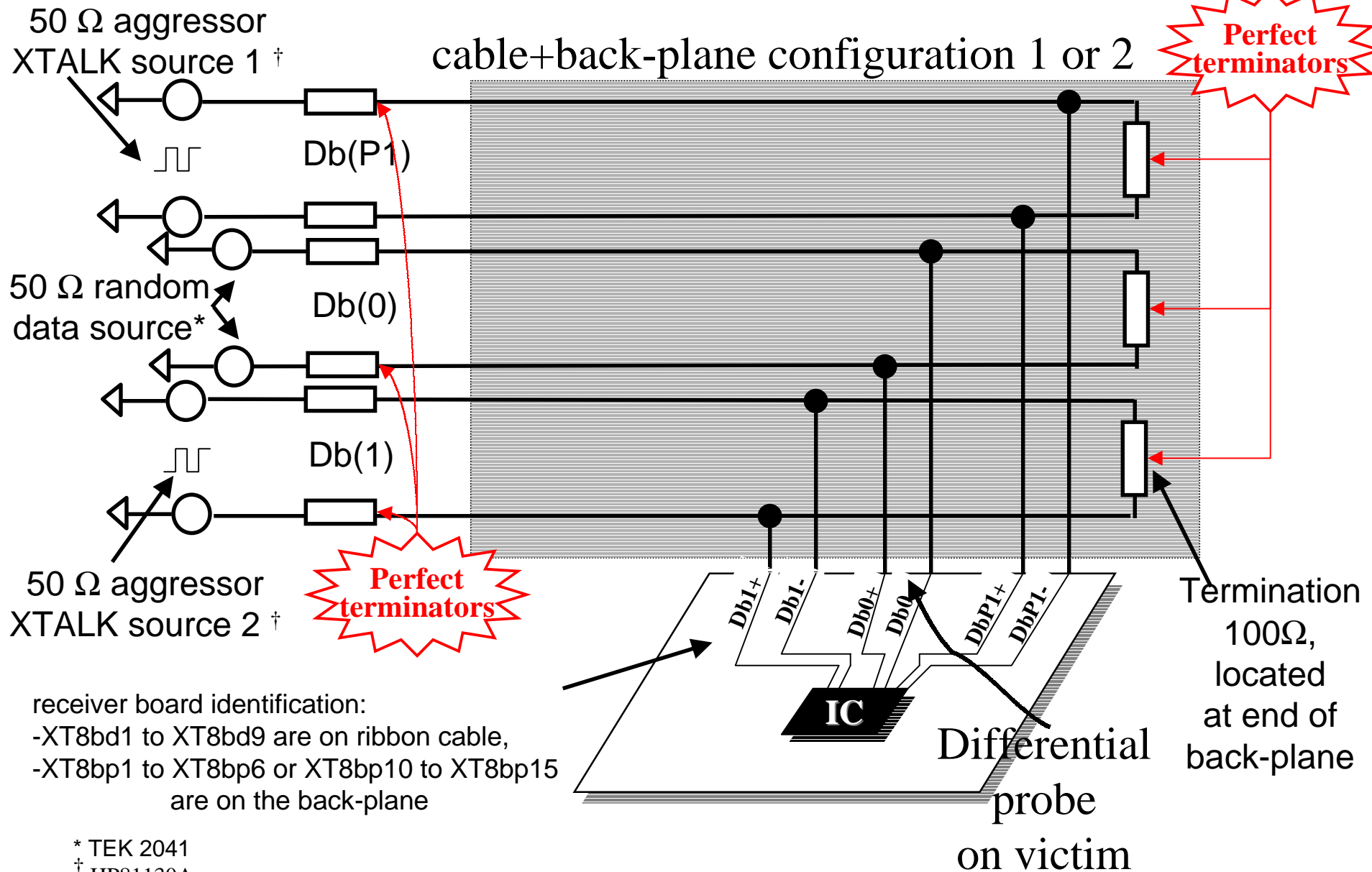
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Config1: Long cable + back-plane

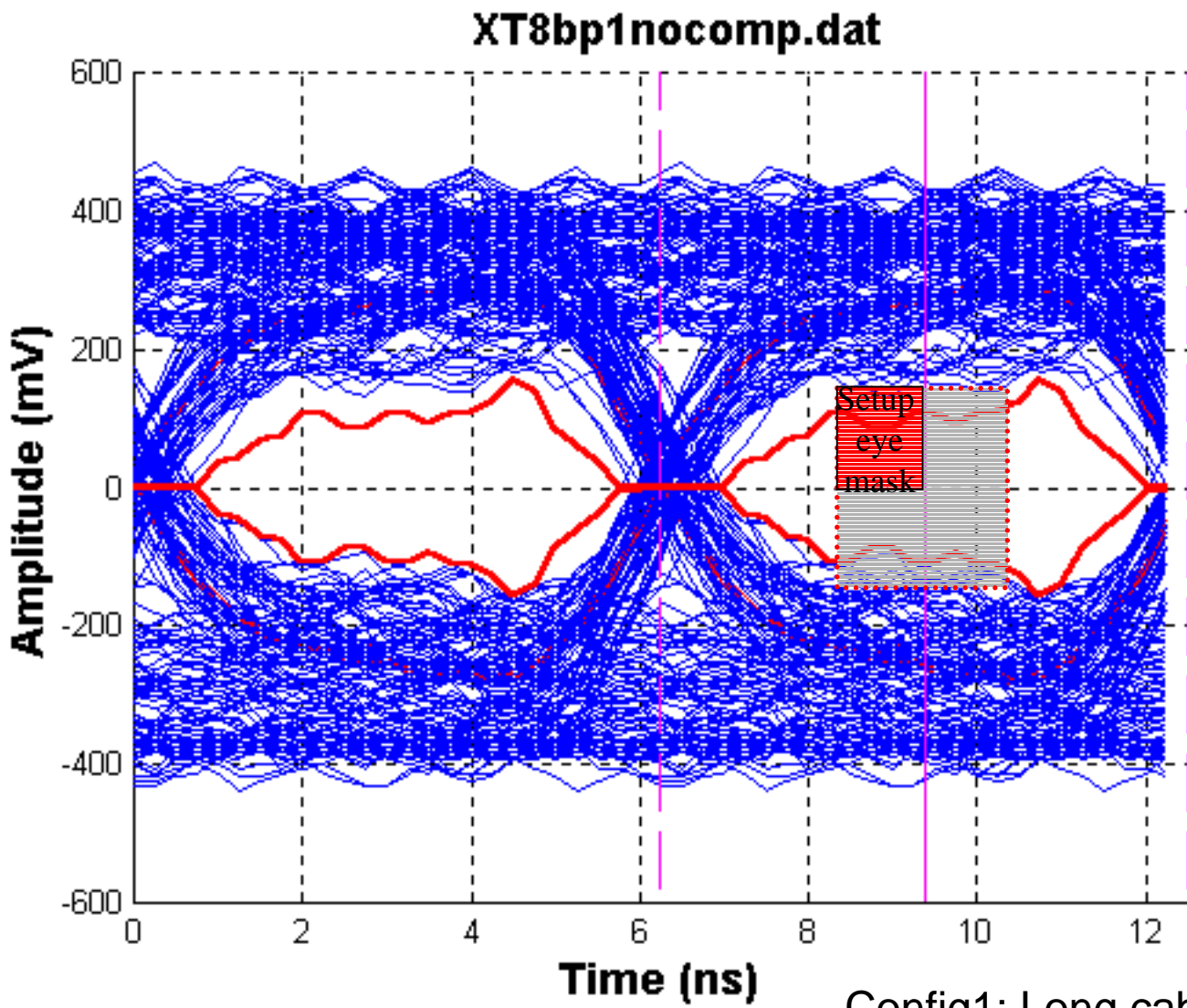


Config2: Short cable + back-plane

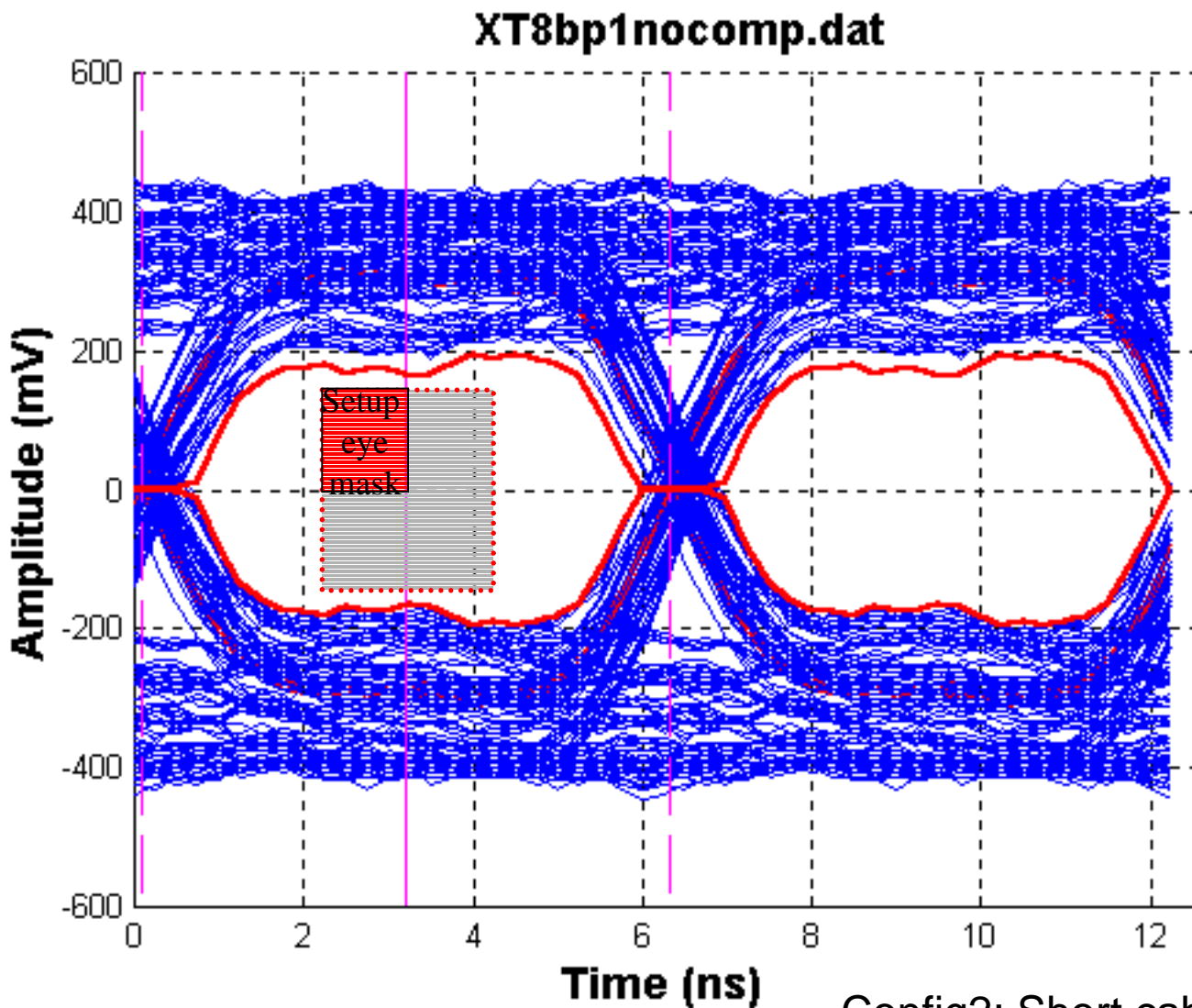


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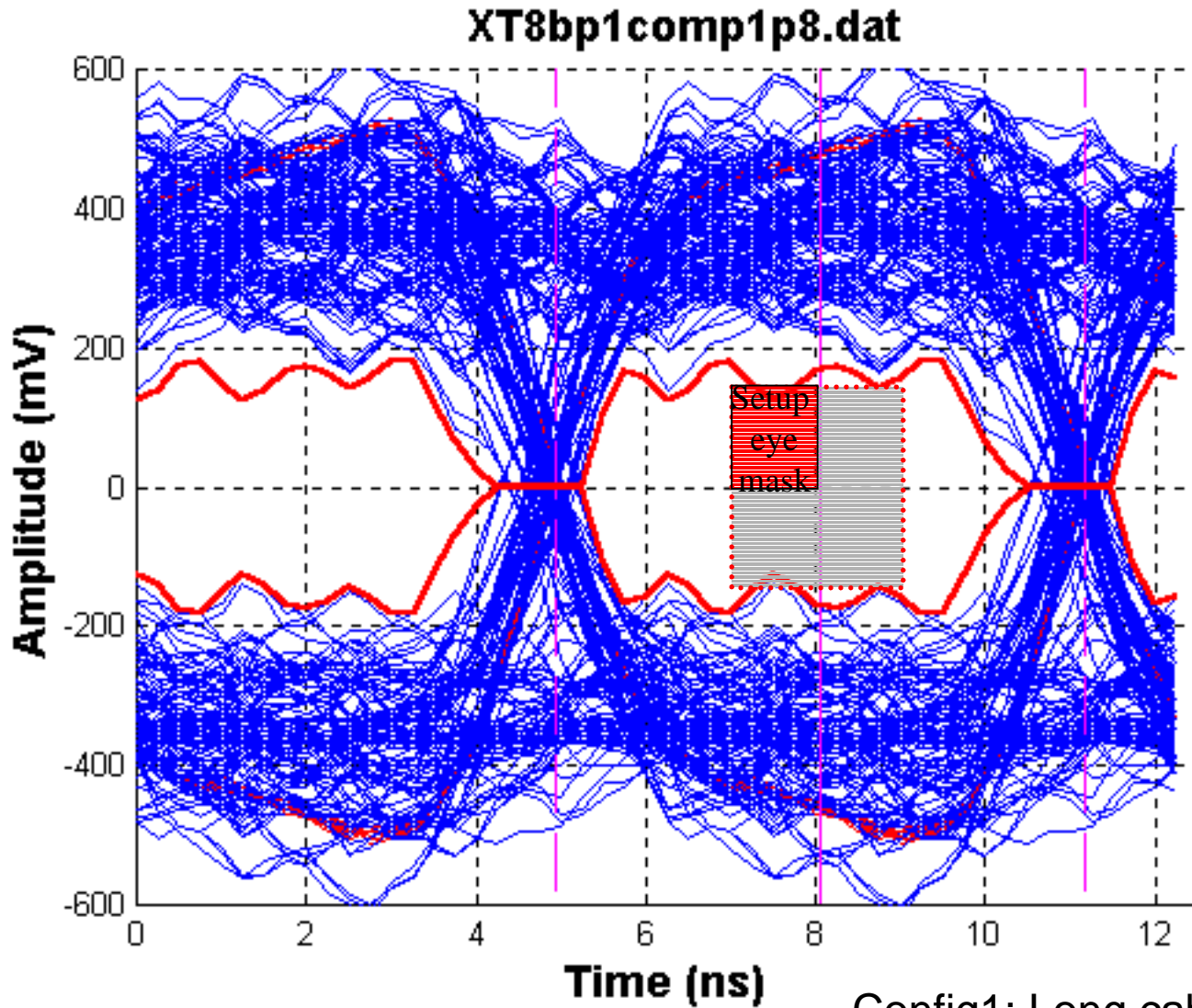
† HP81130A



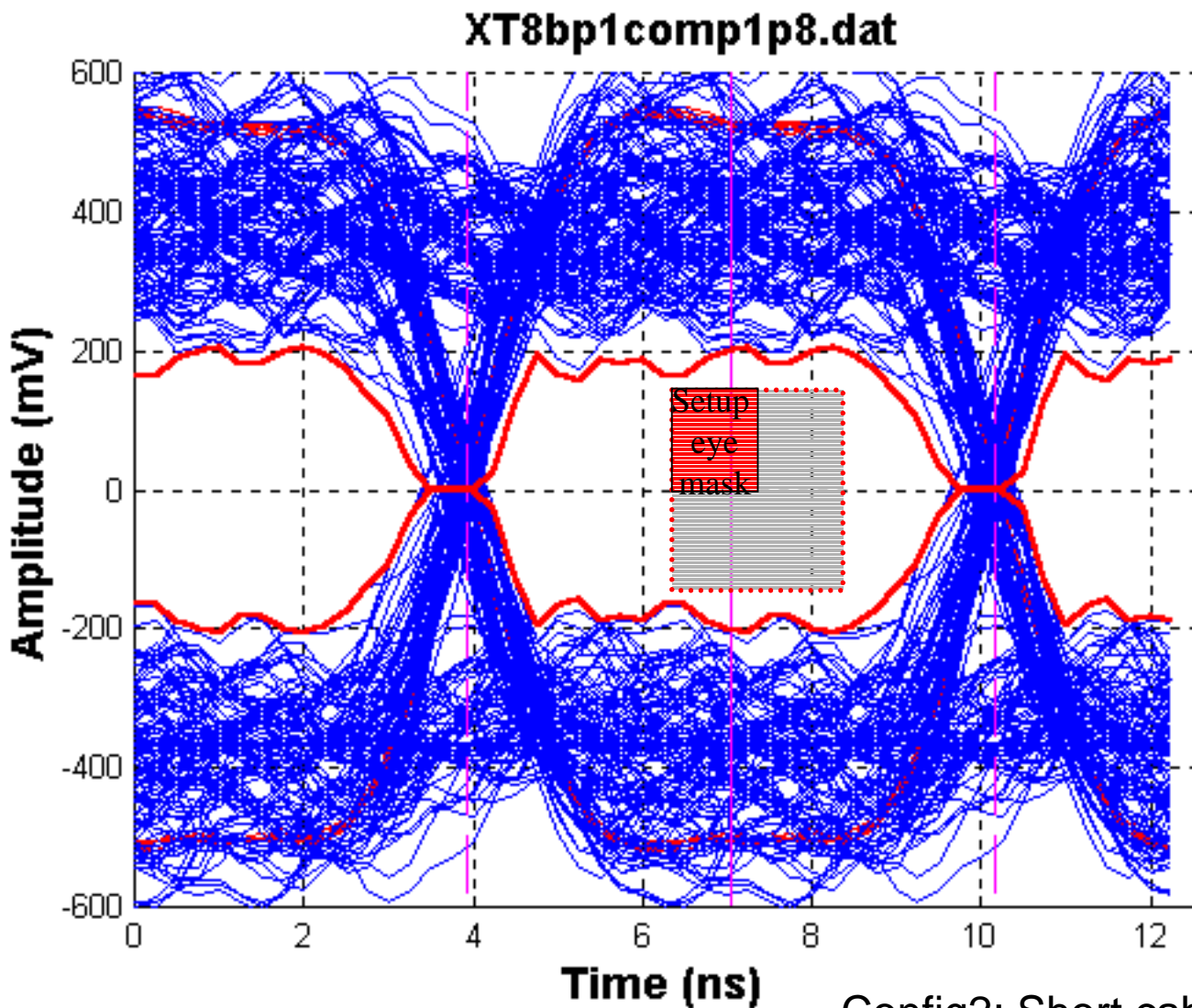
Config1: Long cable + back-plane



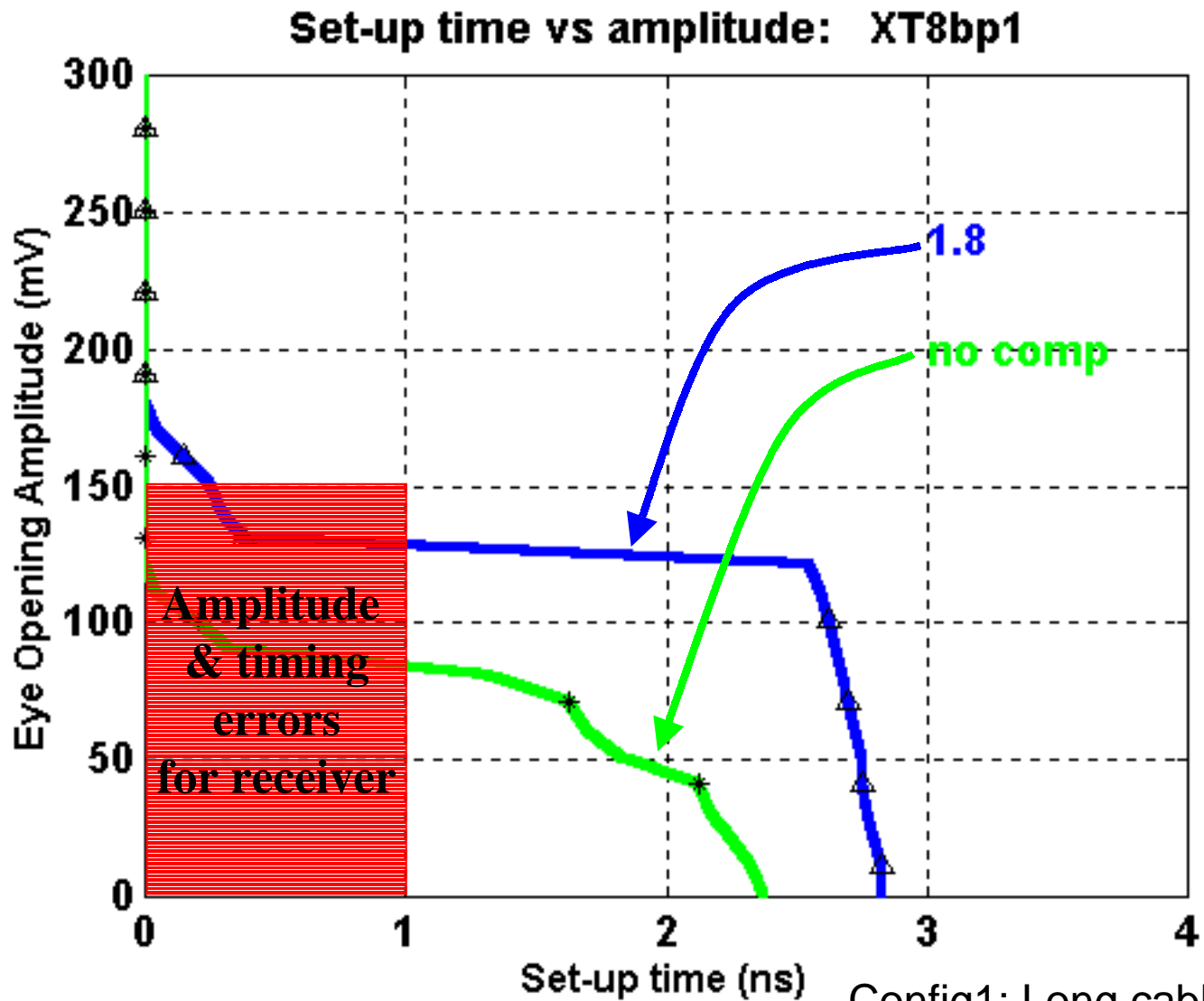
Config2: Short cable + back-plane



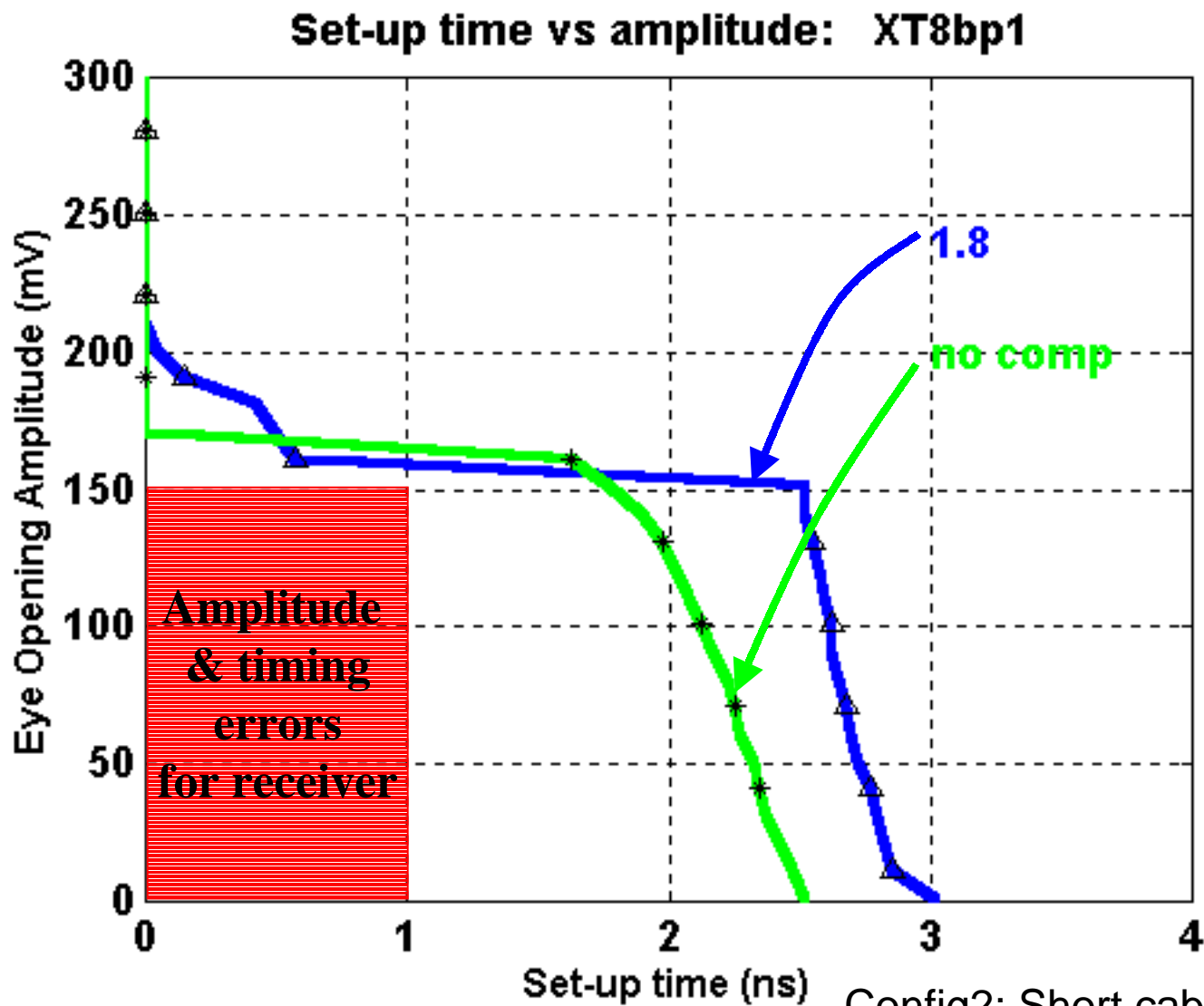
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Config2: Short cable + back-plane



Config1: Long cable + back-plane



Config2: Short cable + back-plane

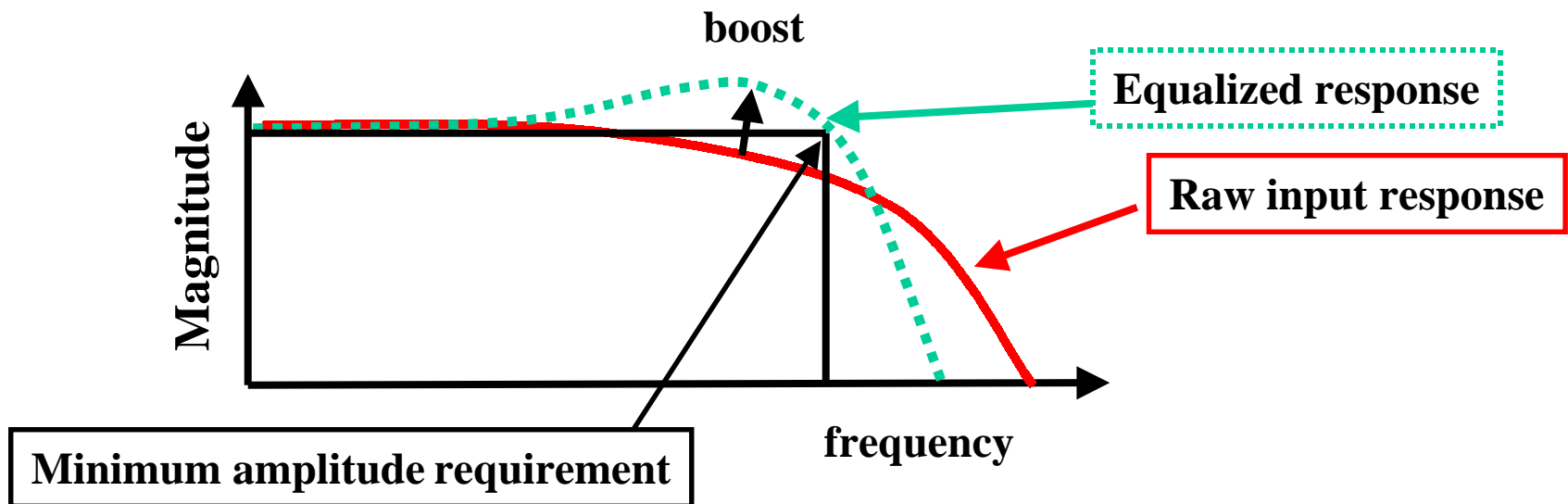
- ① If XTALK is included, transmitter pre-compensation becomes untenable for long configuration:
 - no setup margin
 - no amplitude margin
- ② Also, this data is optimistic:
 - Only two adjacent channels are used for our XTALK measurements,
 - In round cable adjacency relation is unknown; XTALK could be much worse,
 - Sheathed flat cable adjacency is also unknown,
 - Back-plane adjacency is another factor.
- ③ For short flat cable configuration pre-comp does not buy much:
 - slight improvement in set-up margin
 - no improvement in amplitude margin
- ④ We do not believe Pre-comp will work for Ultra-320.

Part II

Receiver Compensation Technique

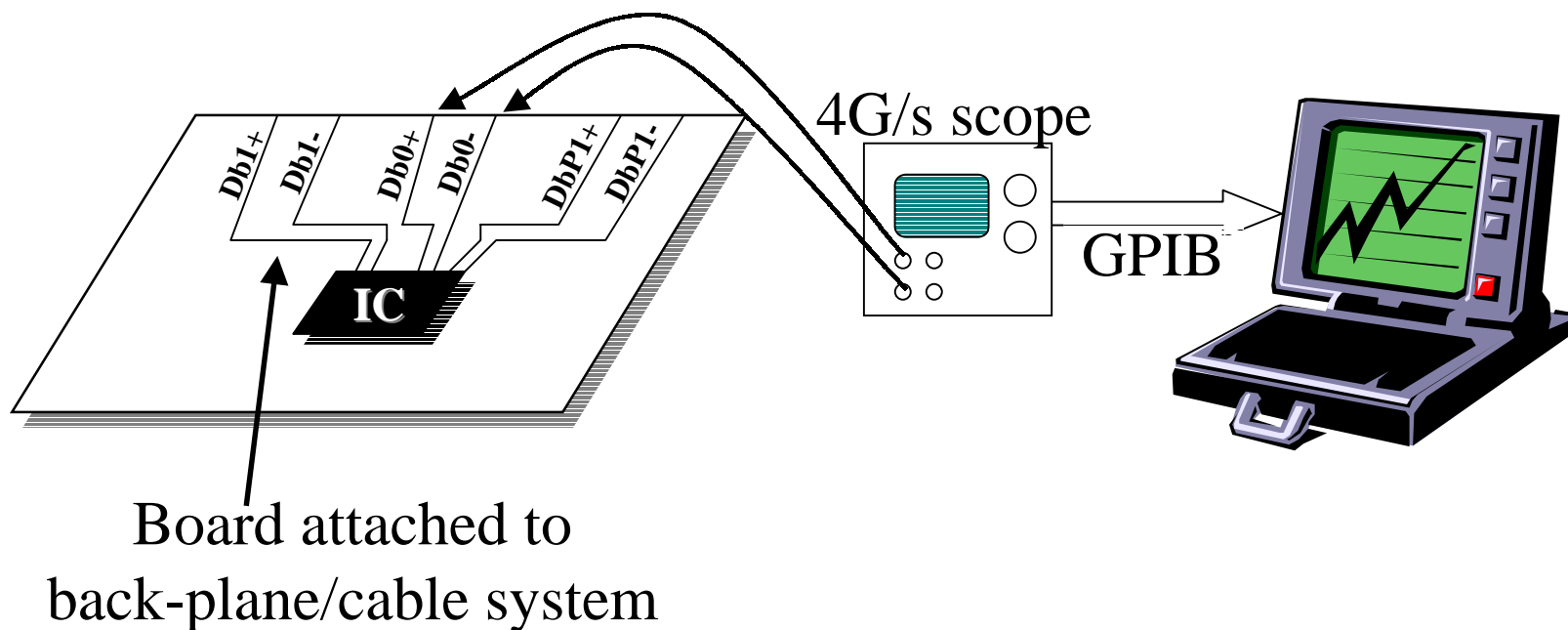
- Receiver equalization overview
- Investigate receiver equalization for Ultra-320 SCSI.
- Demonstrate simulated response to measured data for different configurations.
- Compare receiver eye diagrams for Rx equalizer vs 1.8X pre-comp
 - 400mV peak transmit amplitude used for Rx equalization data
 - 400mV / 720mV transmit amplitudes for 1.8X pre-comp data

- Boost AC magnitude response of cable plant to increase amplitude of isolated pulses. Use high frequency roll-off to minimize noise.

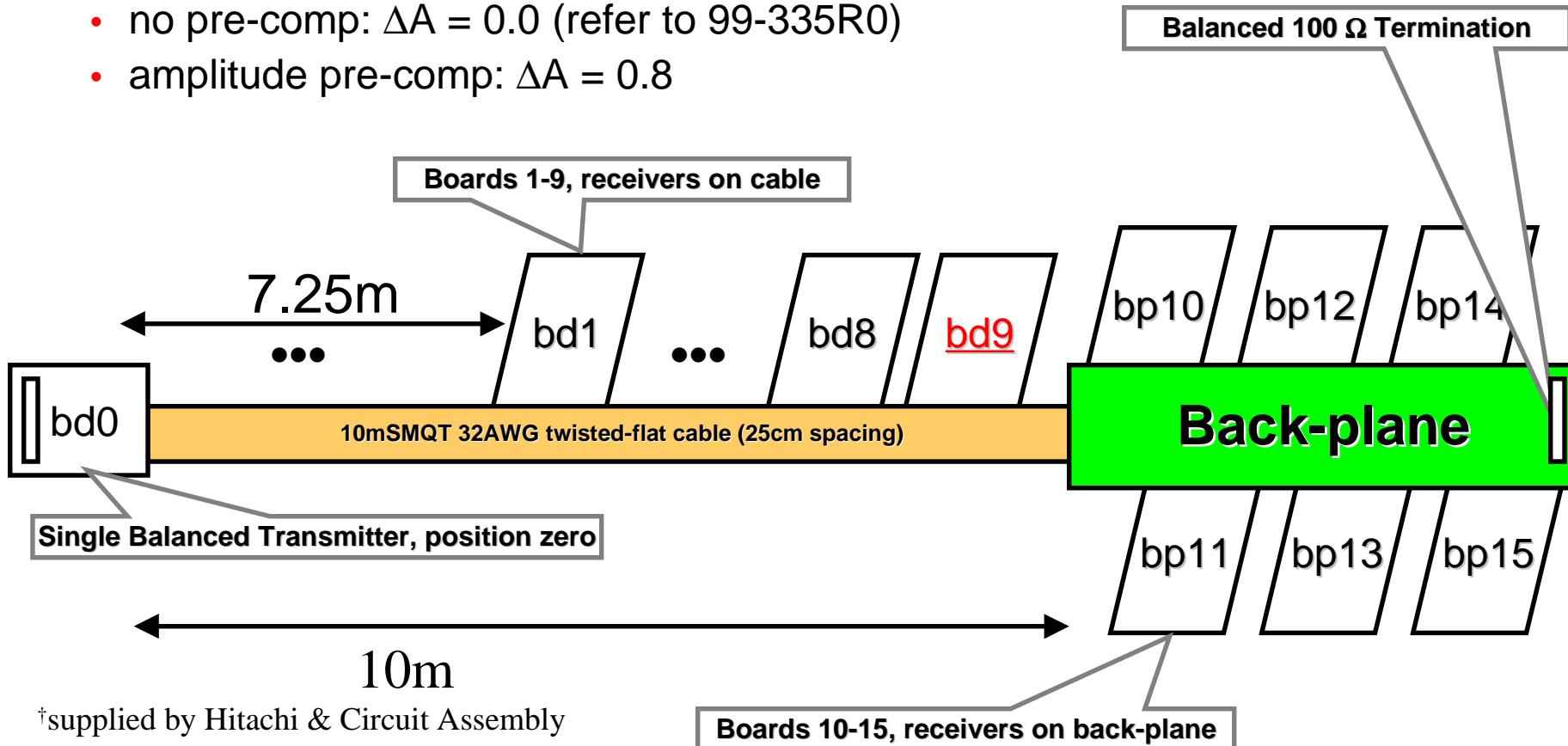


- Need to adapt boost to varying cable conditions, using a pre-defined training pattern so that minimum amplitude response is met for all possible conditions.

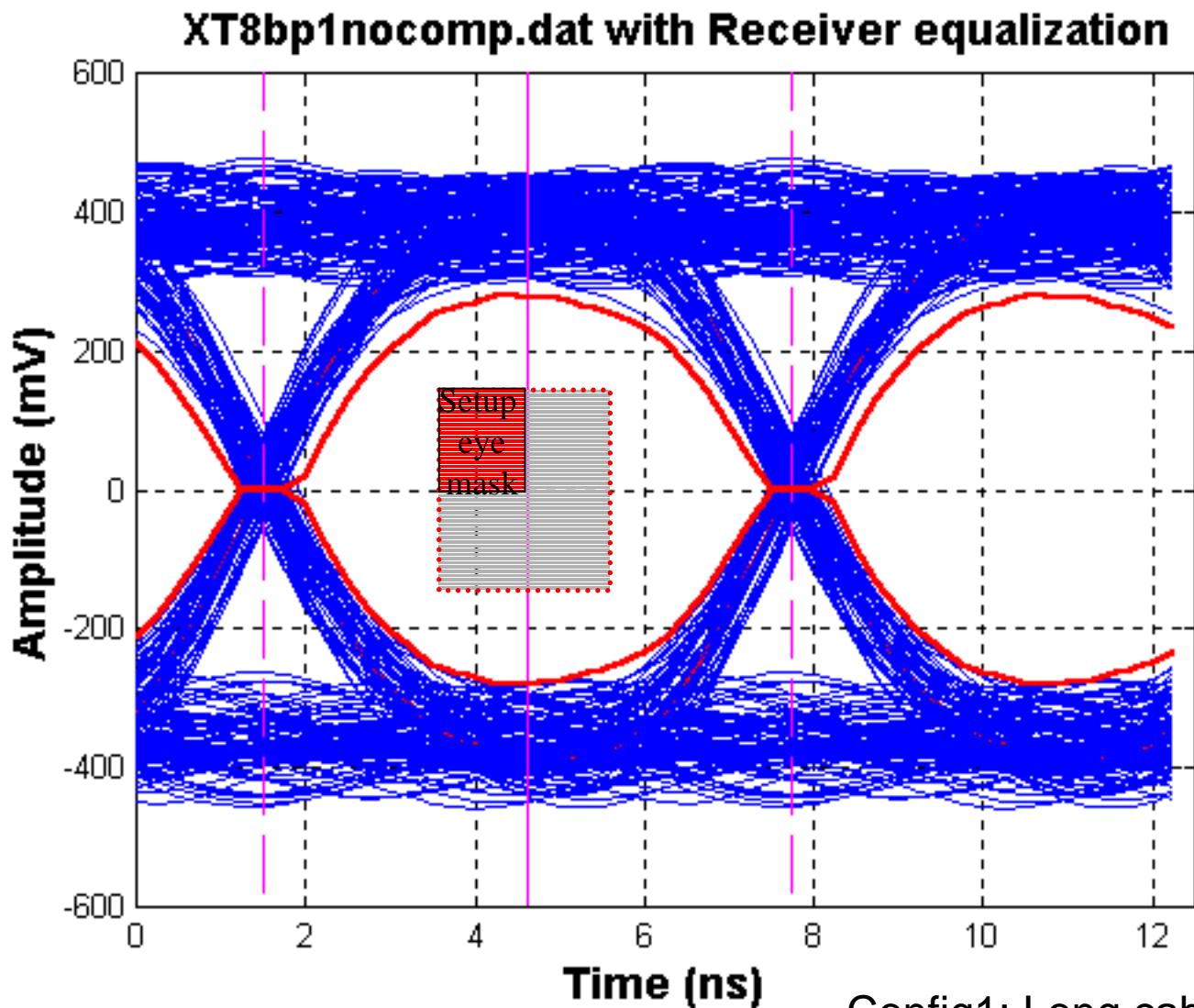
- Capture differential data from a cable and back-plane setup.
- Send raw data to a PC to run mathematical simulation script:
 - same data sets as earlier discussions
 - numerical adaptive equalization
 - evaluate eye diagrams



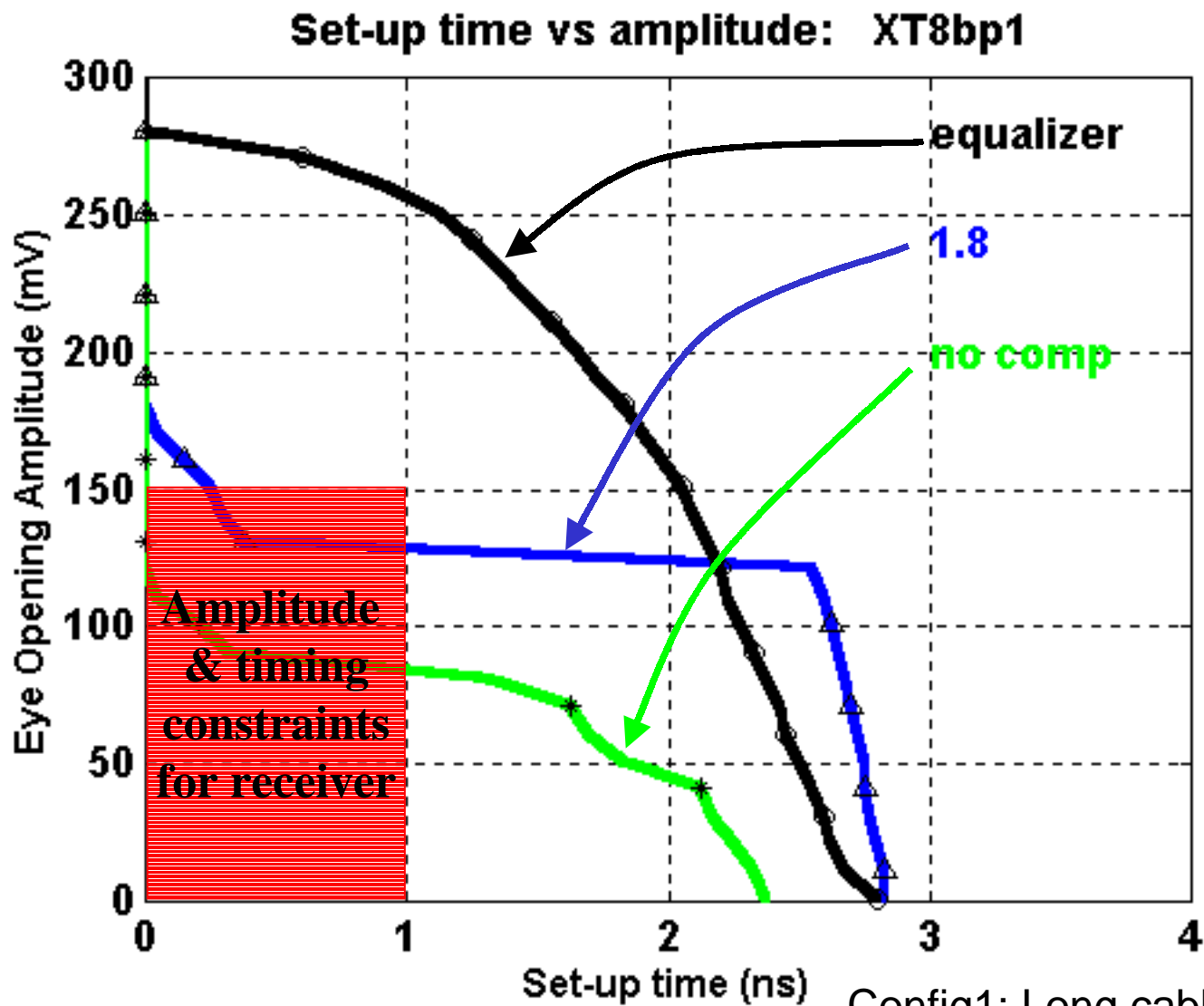
- Hitachi 10 meter, 32AWG twisted-flat ribbon cable[†], 25cm load spacing, plus 6-slot back-plane.
- Waveforms captured @ 4Gs/s:
 - no pre-comp: $\Delta A = 0.0$ (refer to 99-335R0)
 - amplitude pre-comp: $\Delta A = 0.8$



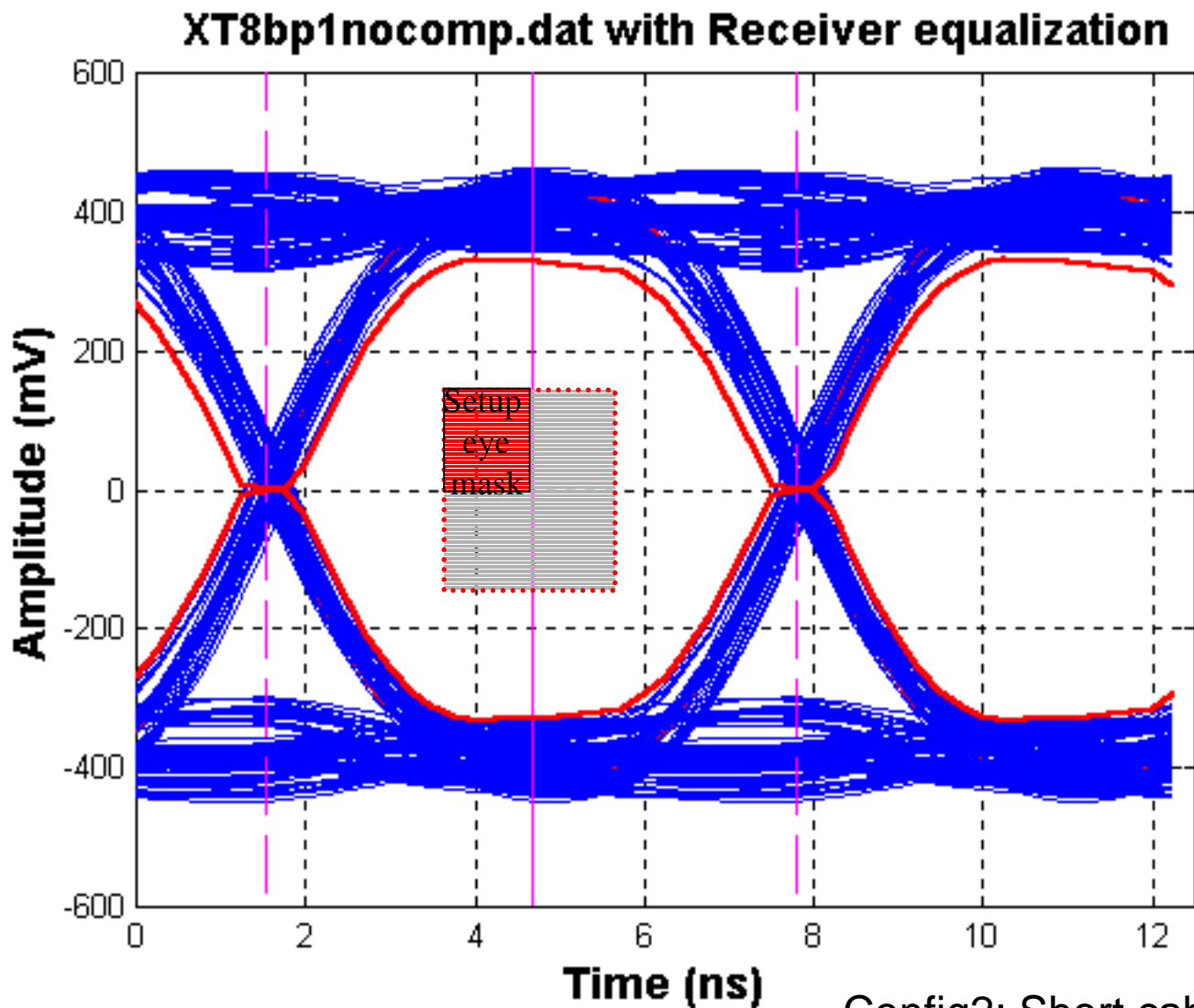
[†]supplied by Hitachi & Circuit Assembly



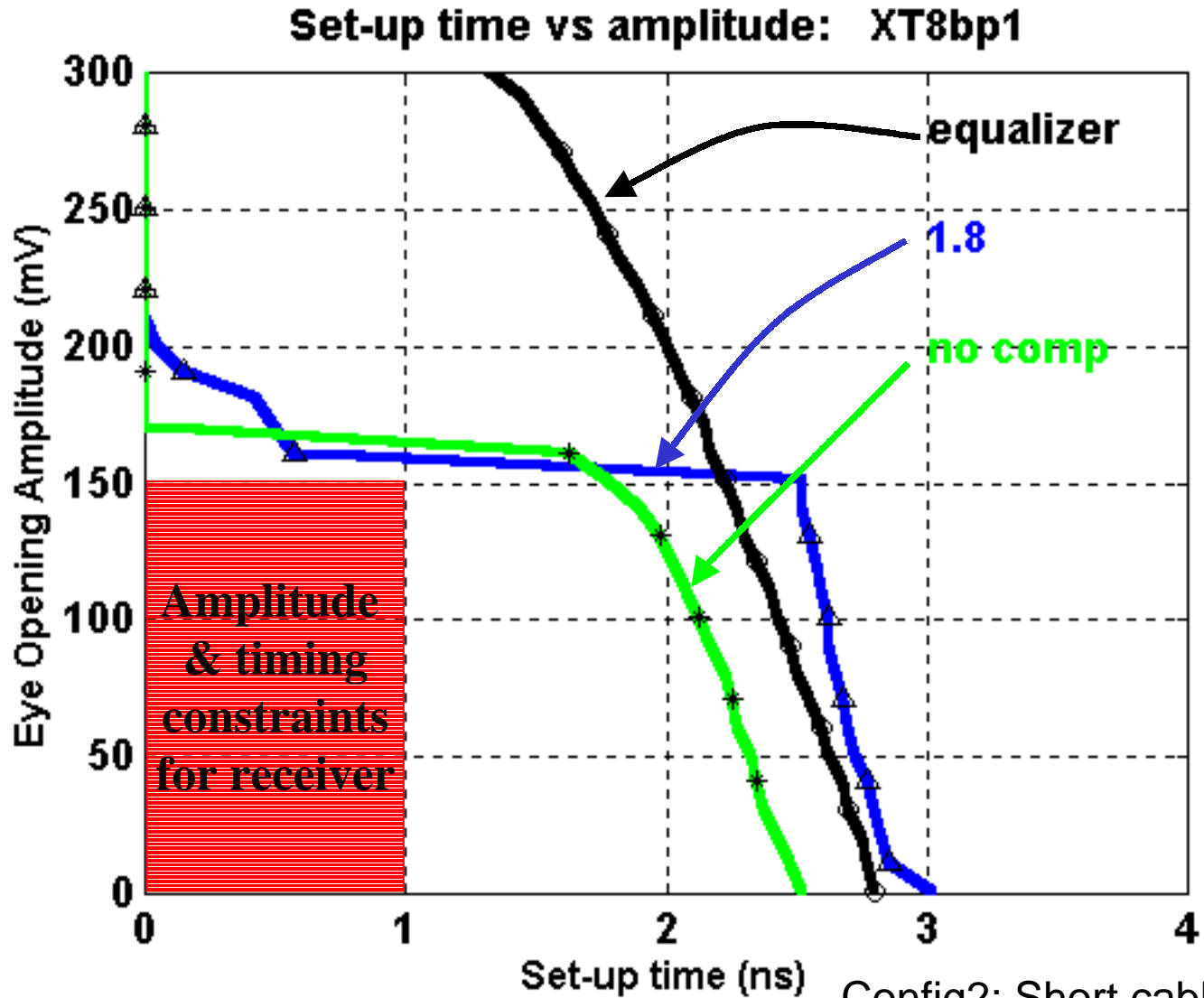
Config1: Long cable + back-plane



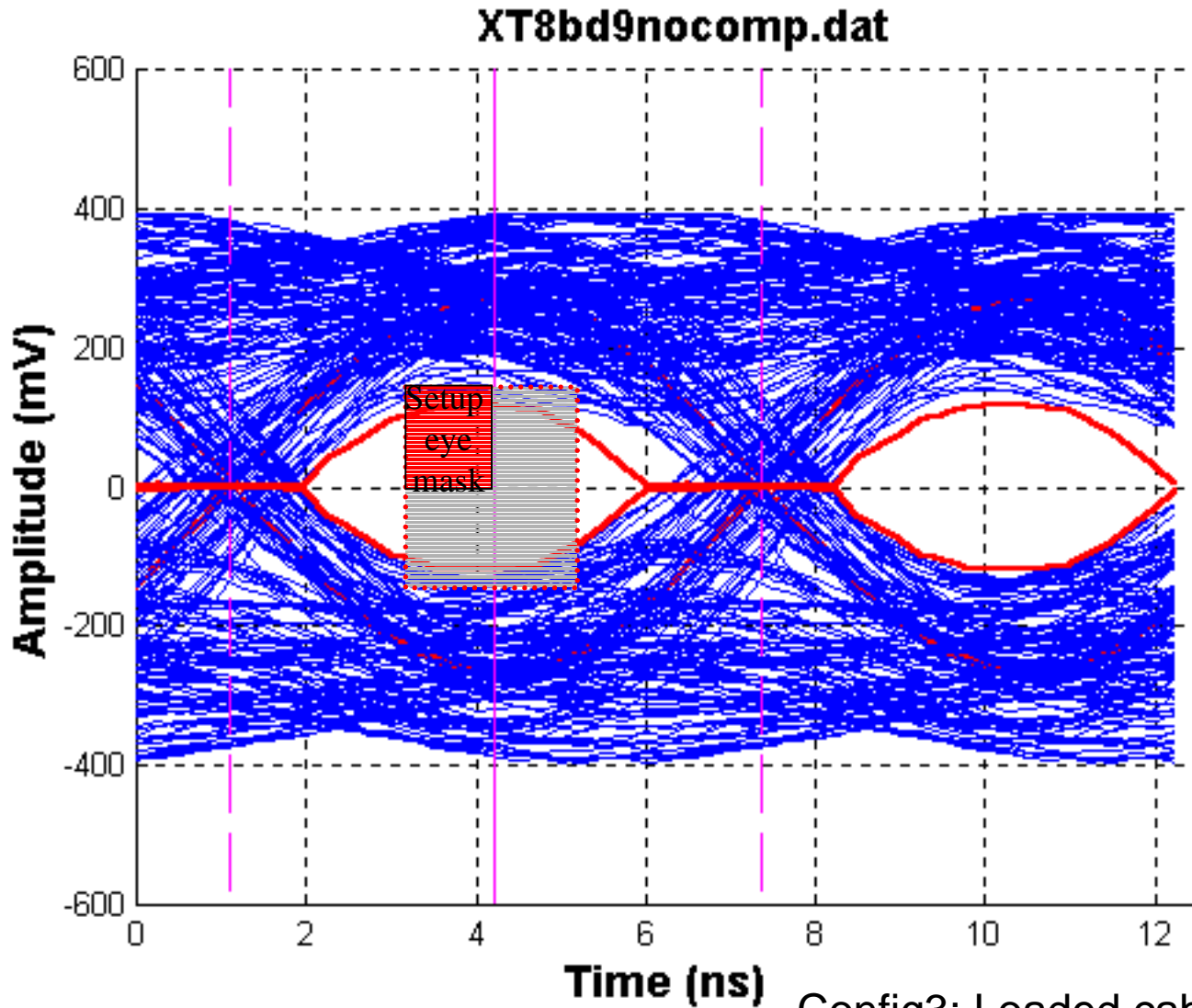
Config1: Long cable + back-plane



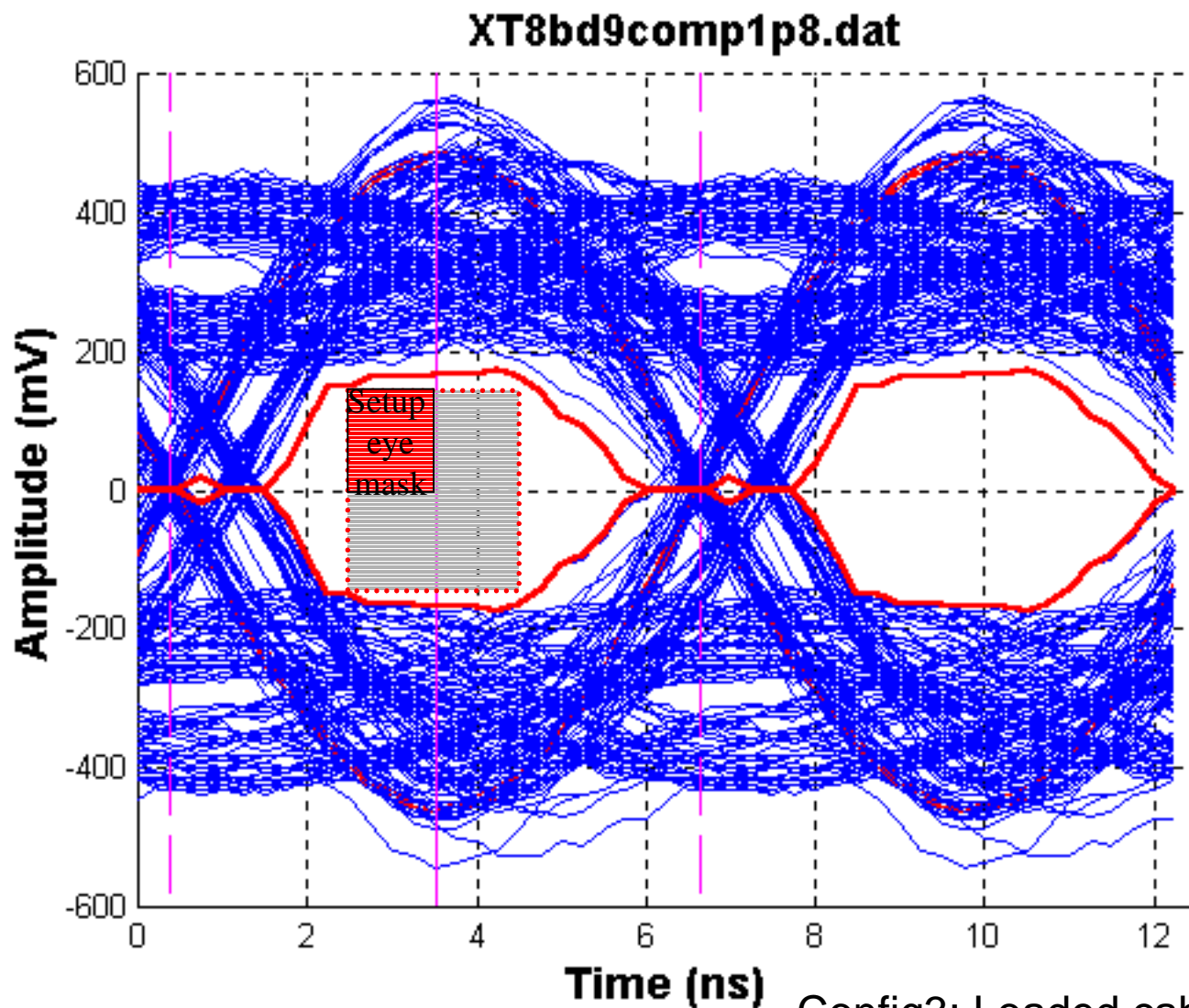
Config2: Short cable + back-plane

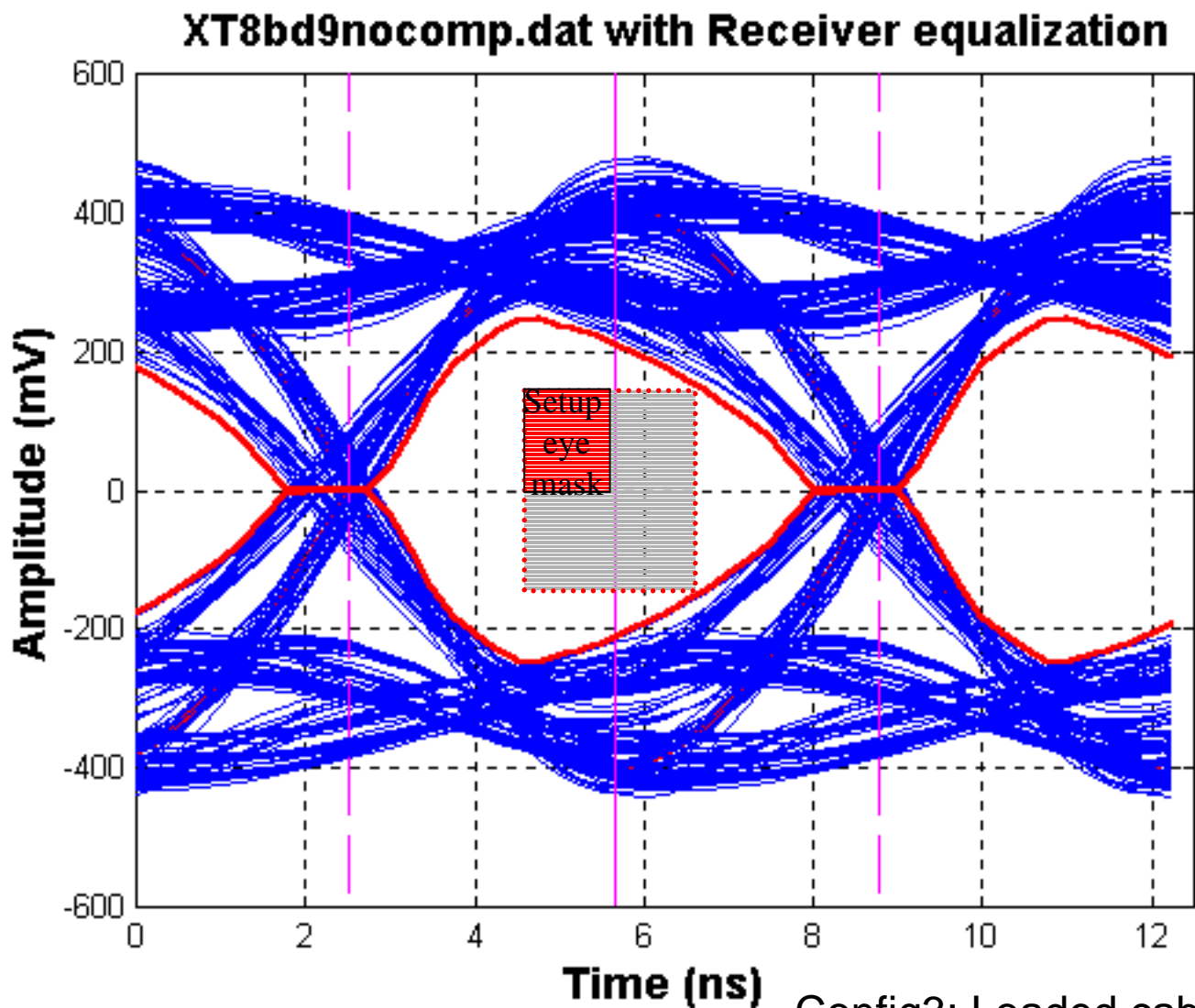


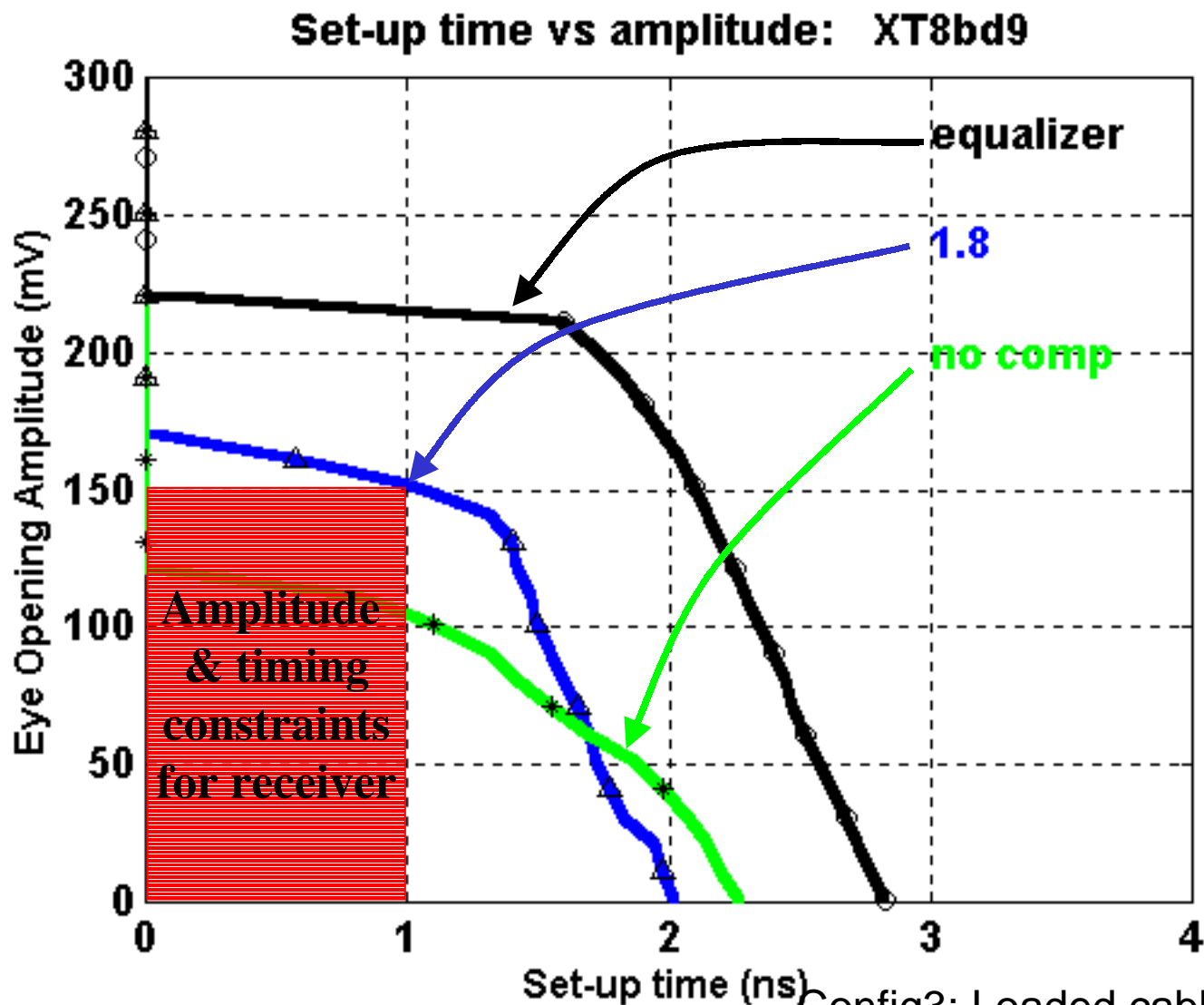
Config2: Short cable + back-plane



Config3: Loaded cable + back-plane







Config3: Loaded cable + back-plane

- ① Receiver equalization boosts performance in all cases
- ② More setup time
- ③ Performance improvements without boosting transmitted signal amplitude
 - no increase in power or transmitter output stage complexity
 - Rx Equalized Data is for a 400mV transmit level, leaving the option of higher transmit levels for improved margins if required.
 - 1.8x pre-comp transmit level is 720mV, leaving no room for improvement.
- ④ Implemented with a “simple adaptive analog” algorithm.
- ⑤ Adapts on a simple training pattern.
- ⑥ Optimum equalization for each bus receiver